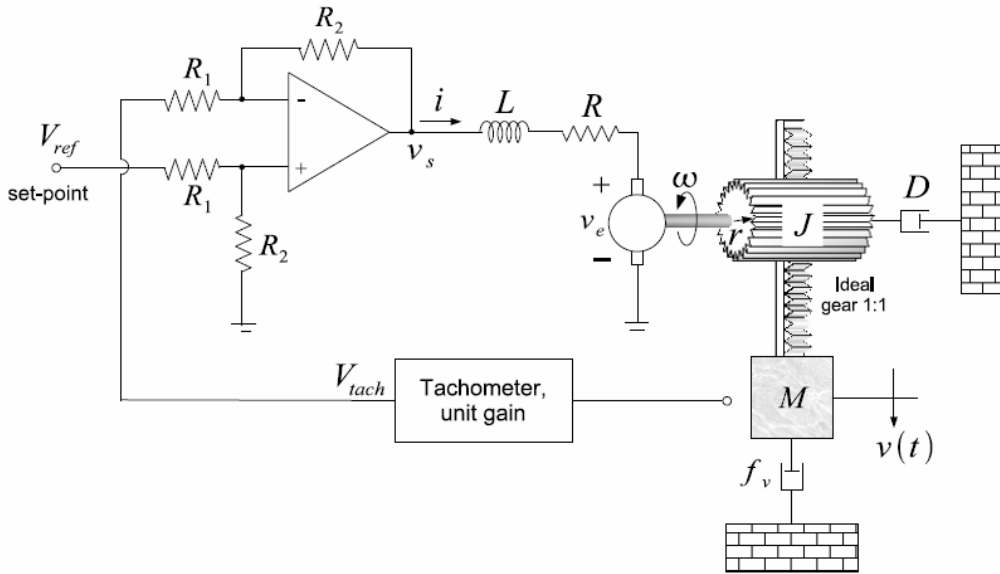


# Last week: analysis of pinion-rack w velocity feedback



Calculation of the steady-state error

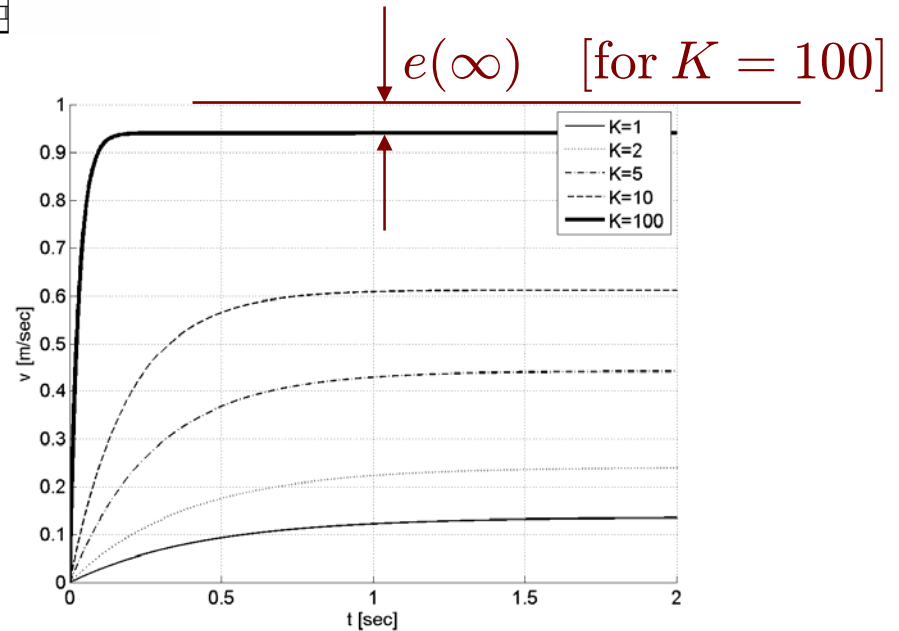
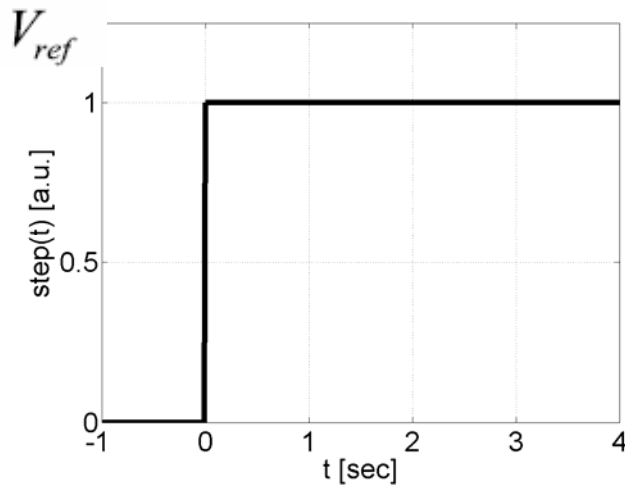
$$\text{Transfer function: } \frac{V(s)}{V_{ref}(s)} = \frac{0.3162K}{s + 2 + 0.3162K}$$

$$\text{Step input: } V_{ref}(s) = \frac{1}{s}$$

$$\text{Output: } V(s) = \frac{1}{s} \times \frac{0.3162K}{s + 2 + 0.3162K}$$

$$\text{Steady-state: } v(\infty) = \lim_{s \rightarrow 0} sV(s) = \frac{0.3162K}{2 + 0.3162K}$$

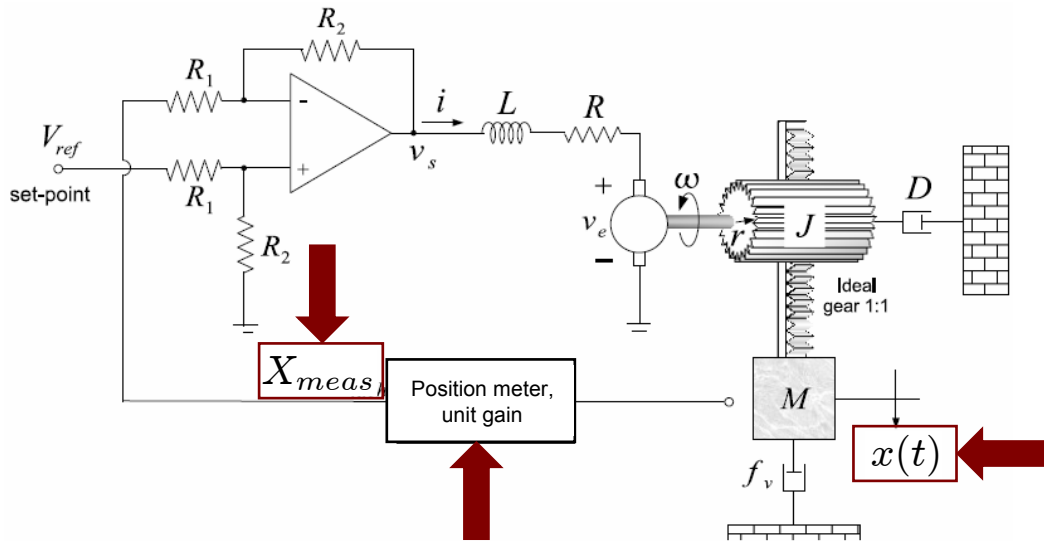
$$\text{Steady-state error: } e(\infty) = 1 - v(\infty) = \frac{2}{2 + 0.3162K}$$



# Today

- Analysis of steady-state errors
  - Definition for step, ramp, parabola inputs
  - Steady-state error in unity feedback systems
  - System type and static error constants
  - The role of integrators
  - Steady-state error in the presence of disturbances
  - Controller gain and step disturbance cancellation
- **Wednesday & Friday:** Root locus

# Inserting an integrator



**We've changed our mind!**  
Now the output is the rack position  $x(t)$ .

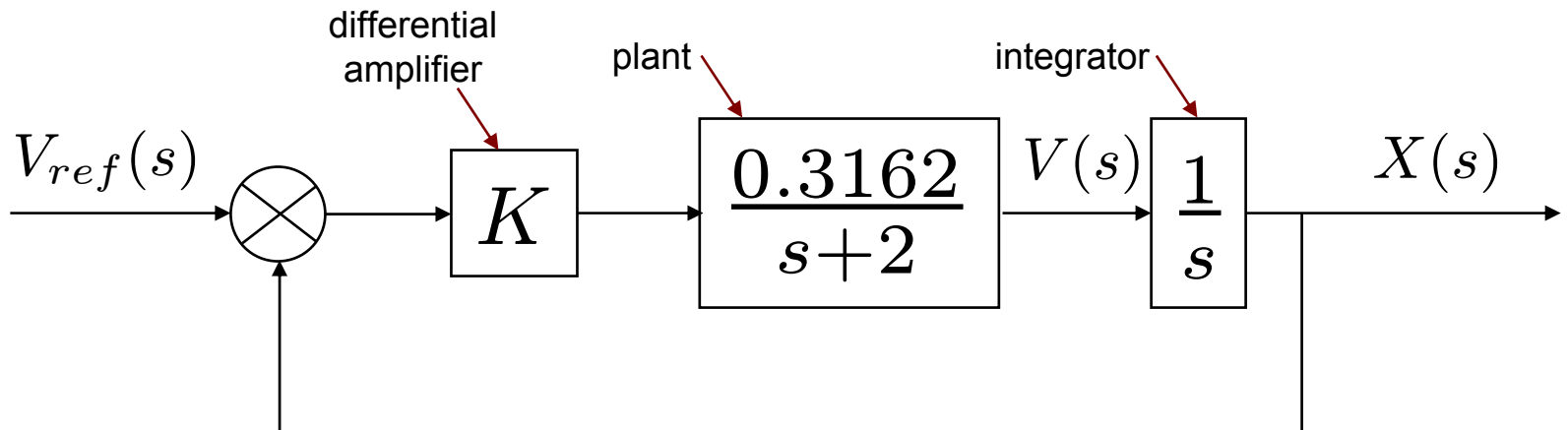
The feedback to the diff-amp must also change to measure position

$$\text{Since } v(t) = \frac{dx(t)}{dt} \Leftrightarrow x(t) = \int_0^t v(t') dt'$$

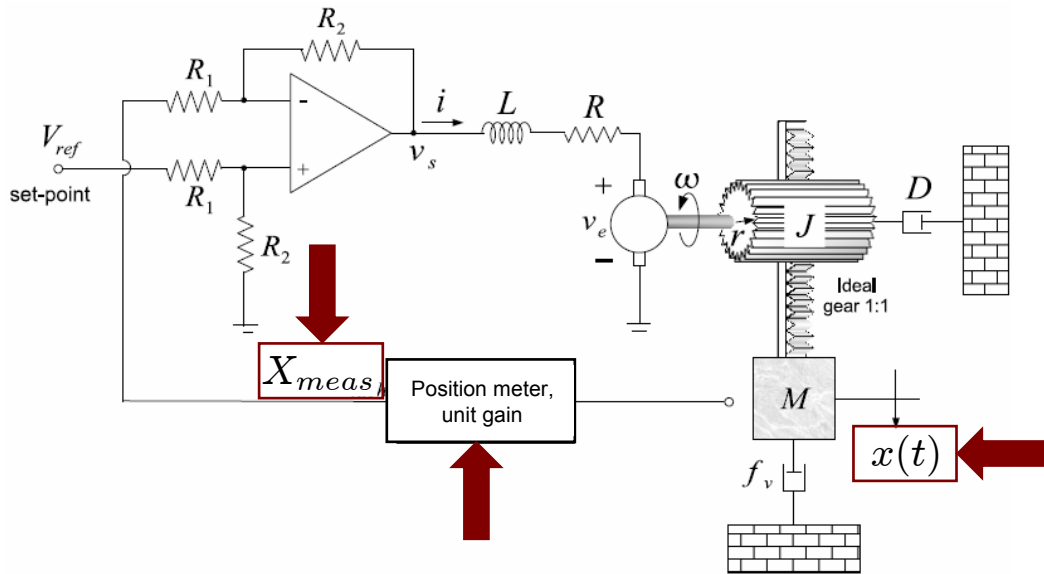
we don't need to re-compute the plant TF.

$$X(s) = \frac{1}{s} V(s),$$

where  $V(s)$  is the output velocity of the previous system that we have already analyzed.



# Inserting an integrator



Calculation of the closed-loop TF

$$\text{Plant TF: } \frac{X(s)}{V_s(s)} = \frac{0.3162}{s(s+2)}$$

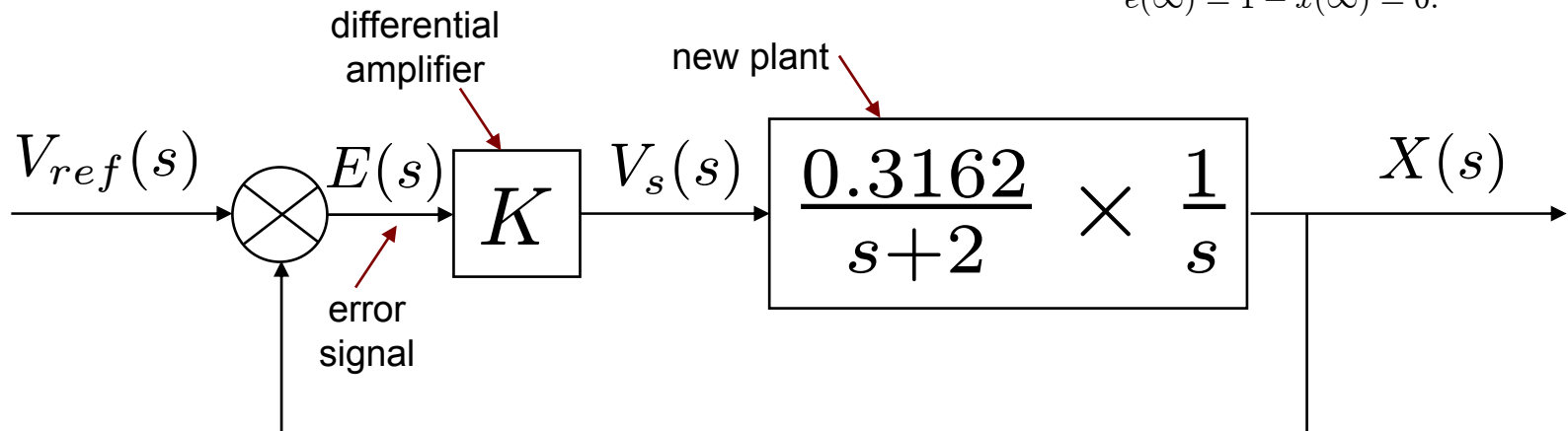
$$\text{Closed loop TF: } \frac{X(s)}{V_{ref}(s)} = \frac{0.3162K}{s^2 + 2s + 0.3162K}$$

Calculation of the steady-state error

$$\text{Step input: } V_{ref}(s) = \frac{1}{s}$$

$$x(\infty) = \lim_{s \rightarrow 0} s \times \frac{1}{s} \times \frac{0.3162K}{s^2 + 2s + 0.3162K} = 1.$$

$$e(\infty) = 1 - x(\infty) = 0.$$



# Generalizing: different system inputs

Images removed due to copyright restrictions.

Please see: Table 7.1 and Fig. 7.1 in Nise, Norman S. *Control Systems Engineering*. 4th ed. Hoboken, NJ: John Wiley, 2004.

# Generalizing: steady-state error for arbitrary input

Images removed due to copyright restrictions.

Please see Fig. 7.2 in Nise, Norman S. *Control Systems Engineering*. 4th ed. Hoboken, NJ: John Wiley, 2004.

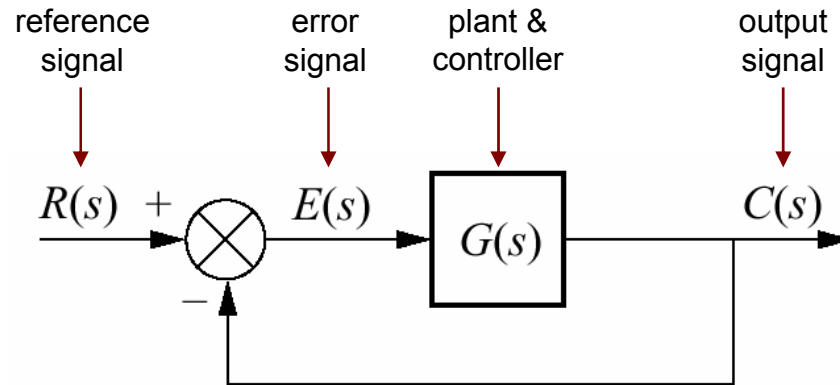
- Unit step input:  
Steady-state error =  
unit step – output as  $t \rightarrow \infty$
- Ramp input:  
Steady-state error =  
ramp – output as  $t \rightarrow \infty$

Generally, the steady-state error is defined as

$$e(\infty) = \lim_{t \rightarrow \infty} [r(t) - c(t)] = \lim_{s \rightarrow 0} s [R(s) - C(s)],$$

where the last equality follows from the final value theorem.

# Generalizing: steady-state error for arbitrary system, unity feedback



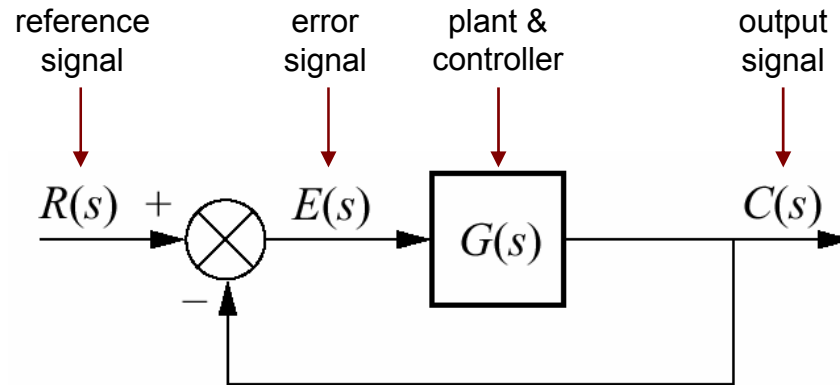
From the definition of the steady-state error,

$$e(\infty) = \lim_{s \rightarrow 0} s \left[ R(s) - C(s) \right] = \lim_{s \rightarrow 0} s E(s).$$

From the block diagram we can also see that

$$\frac{C(s)}{E(s)} = G(s) \Rightarrow E(s) = \frac{C(s)}{G(s)}.$$

# Generalizing: steady-state error for arbitrary system, unity feedback



Recall the closed-loop TF of the unity feedback system

$$\frac{C(s)}{R(s)} = \frac{G(s)}{1 + G(s)} \Rightarrow C(s) = \frac{R(s)G(s)}{1 + G(s)}.$$

Substituting into the two formulae from the previous page,

$$E(s) = \frac{R(s)}{1 + G(s)} \Rightarrow e(\infty) = \lim_{s \rightarrow 0} sE(s) = \lim_{s \rightarrow 0} \frac{sR(s)}{1 + G(s)}.$$



# Steady-state error and static error constants

Image removed due to copyright restrictions.

Please see: Table 7.1 in Nise, Norman S. *Control Systems Engineering*.  
4th ed. Hoboken, NJ: John Wiley, 2004.

$$e(\infty) = \lim_{s \rightarrow 0} \frac{sR(s)}{1 + G(s)}.$$

$$\begin{aligned} e(\infty) &= \lim_{s \rightarrow 0} \frac{1}{s} \times \frac{s}{1+G(s)} \\ &= \lim_{s \rightarrow 0} \frac{1}{1+G(s)} \\ &= \frac{1}{1+\lim_{s \rightarrow 0} G(s)} \equiv \frac{1}{1+K_p} \end{aligned}$$

where  $K_p = \lim_{s \rightarrow 0} G(s)$ .

$$\begin{aligned} e(\infty) &= \lim_{s \rightarrow 0} \frac{1}{s^2} \times \frac{s}{1+G(s)} \\ &= \lim_{s \rightarrow 0} \frac{1}{s+sG(s)} \\ &= \frac{1}{\lim_{s \rightarrow 0} sG(s)} \equiv \frac{1}{1+K_v} \end{aligned}$$

where  $K_v = \lim_{s \rightarrow 0} sG(s)$ .

$$\begin{aligned} e(\infty) &= \lim_{s \rightarrow 0} \frac{1}{s^3} \times \frac{s}{1+G(s)} \\ &= \lim_{s \rightarrow 0} \frac{1}{s^2+s^2G(s)} \\ &= \frac{1}{\lim_{s \rightarrow 0} s^2G(s)} \equiv \frac{1}{1+K_a} \end{aligned}$$

where  $K_a = \lim_{s \rightarrow 0} s^2G(s)$ .

Note: the system must be **stable** (*i.e.*, all poles on left-hand side or at the origin) for these calculations to apply

# System types and steady-state errors

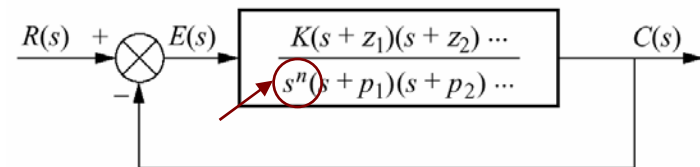
Table removed due to copyright restrictions.

Please see: Table 7.2 in Nise, Norman S. *Control Systems Engineering*. 4th ed. Hoboken, NJ: John Wiley, 2004.

$$\mathcal{L}[u(t)] = \frac{1}{s} \quad K_p = \lim_{s \rightarrow 0} G(s)$$

$$\mathcal{L}[tu(t)] = \frac{1}{s^2} \quad K_v = \lim_{s \rightarrow 0} sG(s)$$

$$\mathcal{L}\left[\frac{1}{2}t^2u(t)\right] = \frac{1}{s^3} \quad K_a = \lim_{s \rightarrow 0} s^2G(s)$$

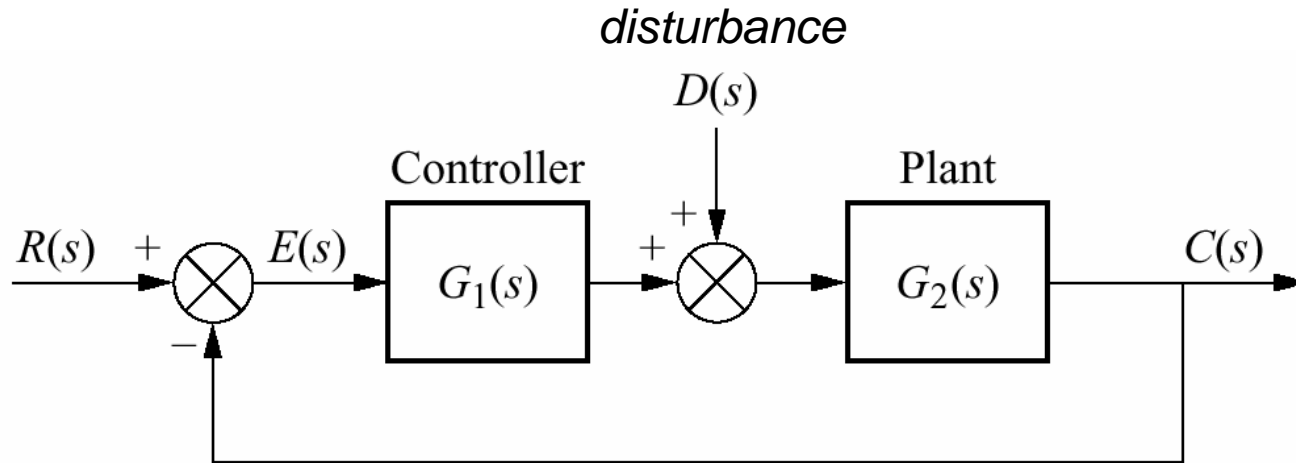


$$n = 0 \quad \text{Type 0}$$

$$n = 1 \quad \text{Type 1}$$

$$n = 2 \quad \text{Type 2}$$

# Disturbances



From the I-O relationship of the plant,

$$C(s) = E(s)G_1(s)G_2(s) + D(s)G_2(s).$$

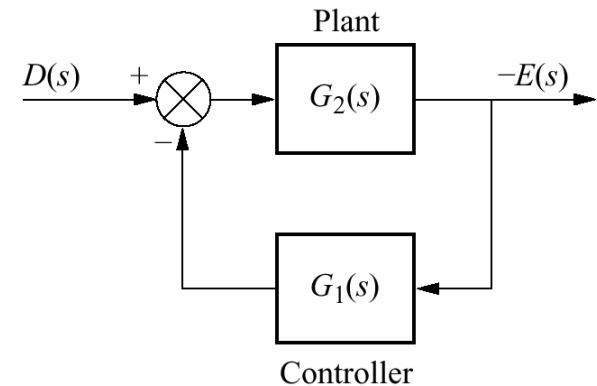
From the summation element,

$$E(s) = R(s) - C(s).$$

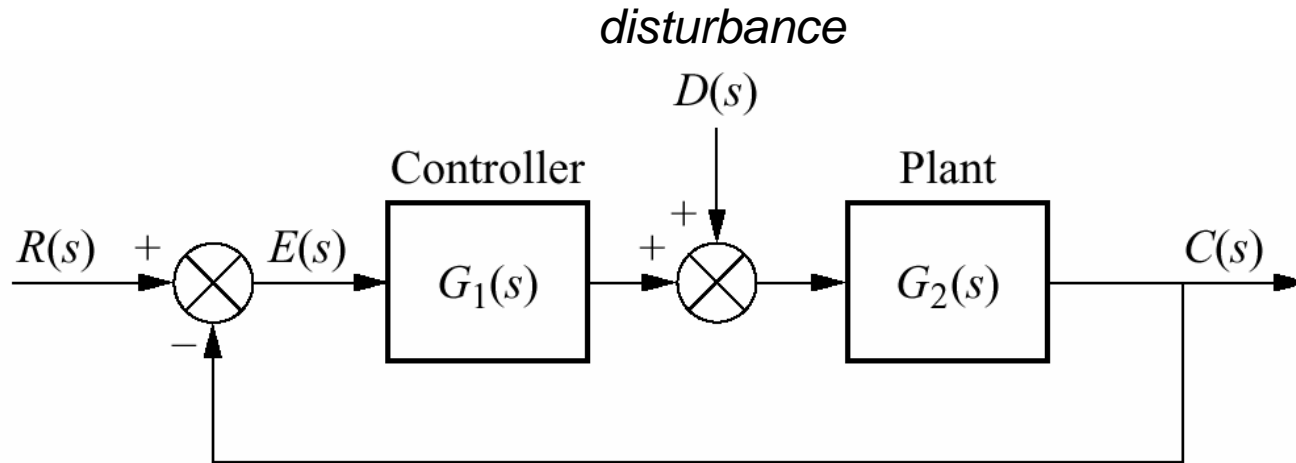
Substituting  $C(s)$  and solving for  $E(s)$ ,

$$E(s) = R(s) \frac{1}{1 + G_1(s)G_2(s)} - D(s) \frac{G_2(s)}{1 + G_1(s)G_2(s)}$$

Equivalent block diagram with  $D(s)$  as input and  $-E(s)$  as output.



# Disturbances

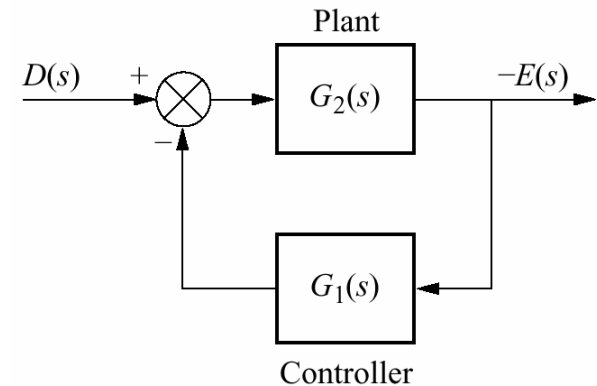


$$e(\infty) = \lim_{s \rightarrow 0} sE(s) = \lim_{s \rightarrow 0} s \left[ R(s) \frac{1}{1 + G_1(s)G_2(s)} - D(s) \frac{G_2(s)}{1 + G_1(s)G_2(s)} \right] \equiv e_R(\infty) + e_D(\infty),$$

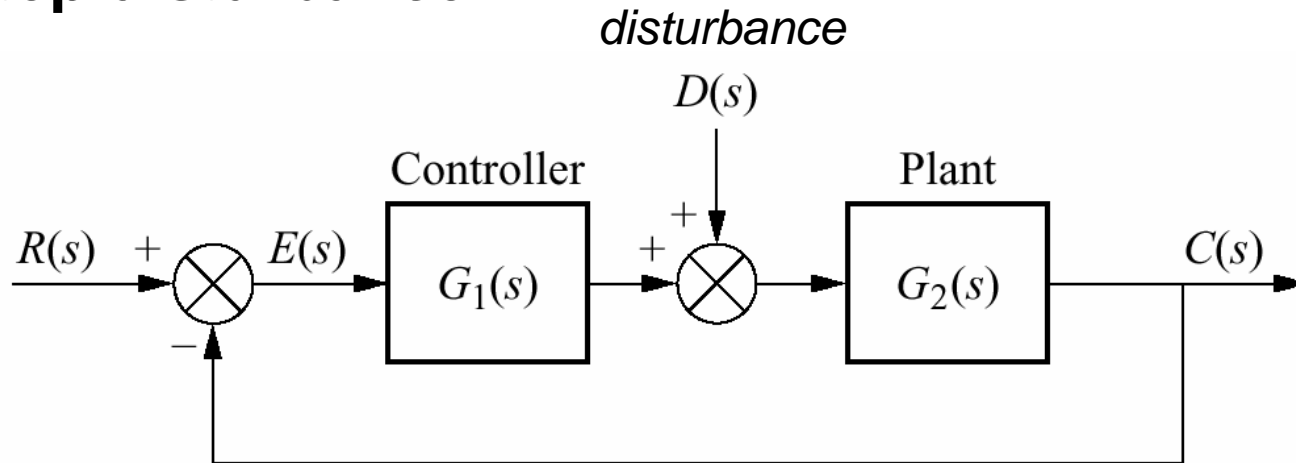
where

$$e_R(\infty) = \lim_{s \rightarrow 0} \frac{sR(s)}{1 + G_1(s)G_2(s)}$$

$$e_D(\infty) = -\lim_{s \rightarrow 0} \frac{sG_2(s)D(s)}{1 + G_1(s)G_2(s)}.$$



# Unit step disturbance

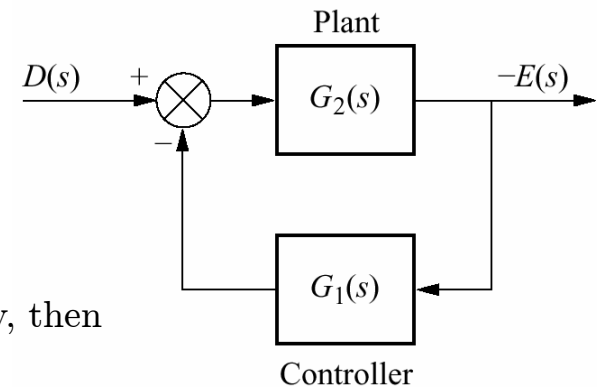


Special case: unit step disturbance  $D(s) = \frac{1}{s}$ .

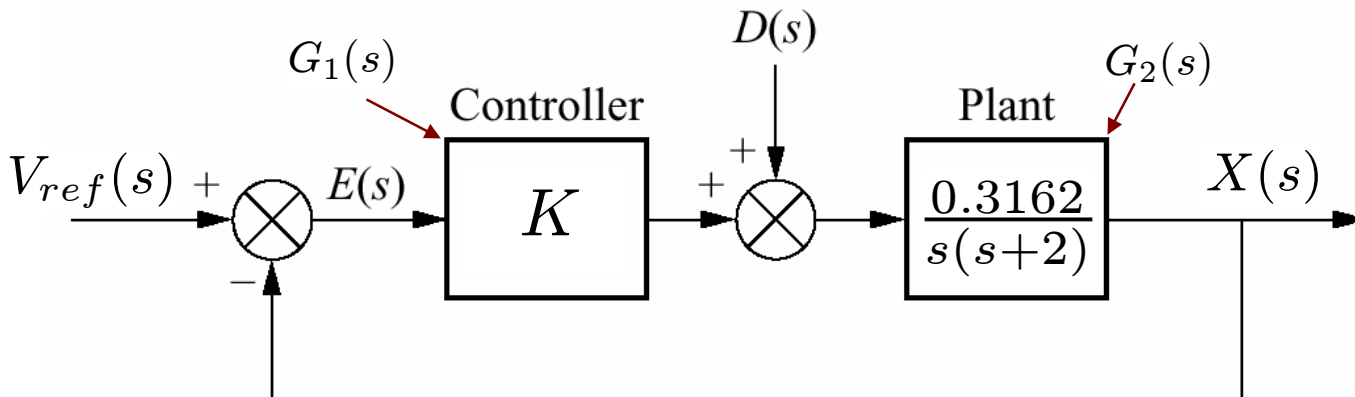
$$\begin{aligned}
 e_D(\infty) &= -\lim_{s \rightarrow 0} \frac{sG_2(s) \times (1/s)}{1 + G_1(s)G_2(s)} \\
 &= -\frac{\lim_{s \rightarrow 0} G_2(s)}{1 + \lim_{s \rightarrow 0} G_1(s)G_2(s)} \\
 &= -\frac{1}{\frac{1}{\lim_{s \rightarrow 0} G_2(s)} + \lim_{s \rightarrow 0} G_1(s)}.
 \end{aligned}$$

If  $K_1$ ,  $K_2$  are the gains of the controller and plant, respectively, then

$$e(\infty) \downarrow \text{ if } K_1 \uparrow \text{ or } K_2 \downarrow.$$



# Unit step disturbance: example



DC motor with rack–pinion load, position feedback,

subject to unit step disturbance  $D(s) = \frac{1}{s}$ .

$$\begin{aligned}
 e_D(\infty) &= - \frac{1}{\frac{1}{\lim_{s \rightarrow 0} G_2(s)} + \lim_{s \rightarrow 0} G_1(s)} \\
 &= - \frac{1}{\lim_{s \rightarrow 0} \frac{s(s+2)}{0.3162} + \lim_{s \rightarrow 0} K} \\
 &= - \frac{1}{K}.
 \end{aligned}$$

