

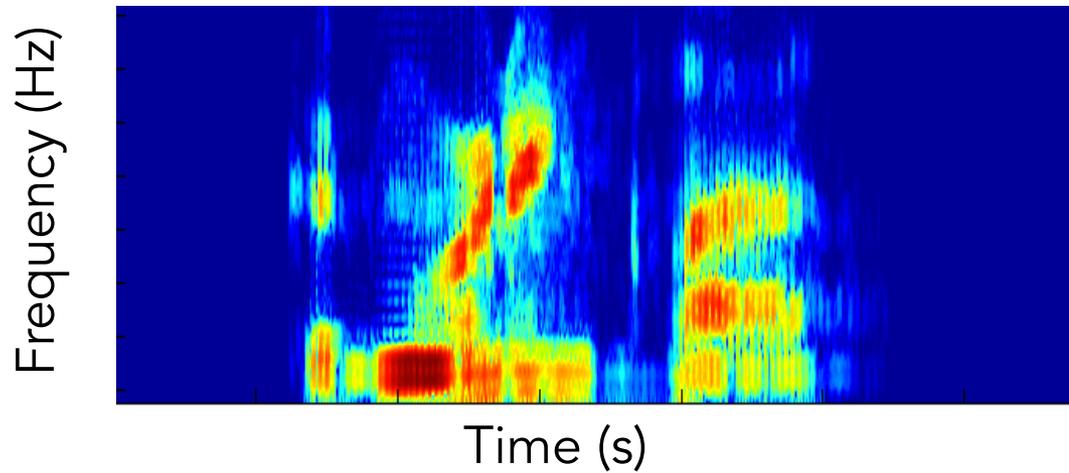
# Introduction to Neural Computation

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Prof. Michale Fee  
MIT BCS 9.40 — 2018

Lecture 11 - Spectral analysis I

# Spectral Analysis



Game plan for Lectures 11, 12, and 13 —  
*Develop a powerful set of methods for  
understanding the temporal structure of signals*

- Fourier series, Complex Fourier series, Fourier transform, Discrete Fourier transform (DFT), Power Spectrum
- Convolution Theorem
- Noise and Filtering
- Shannon-Nyquist Sampling Theorem
  - <https://markusmeister.com/2018/03/20/death-of-the-sampling-theorem/>
- Spectral Estimation
- Spectrograms
- Windowing, Tapers, and Time-Bandwidth Product
- Advanced Filtering Methods

# Learning Objectives for Lecture 11

- Fourier series for symmetric and asymmetric functions
- Complex Fourier series
- Fourier transform
- Discrete Fourier transform (Fast Fourier Transform - FFT)
- Power spectrum

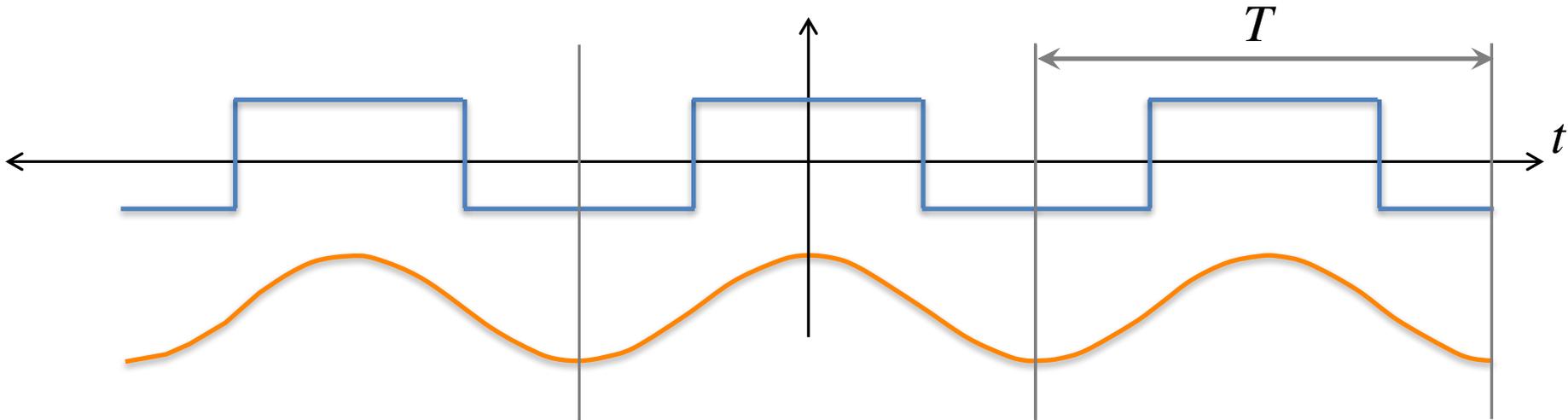
# Discrete Fourier transform

- Some code

```
WSpec.m x recordaudio.m x continuous_cos.m x +
1 %
2 - N=2048; % number of samples in time
3 %
4 - dt=.001; % sampling interval
5 - Fs=1./dt; % sampling frequency
6 - time=dt*[-N/2:N/2-1]; % timebase
7 %
8 - freq=20.; % frequency of sine wave in Hz
9 - y=cos(2*pi*freq*time);
10 %
11 %
12 - yshft=circshift(y,[0,N/2]); % First shift zero point from center to
13 % first point in the array
14 - ffty=fft(yshft, N)/N; % Now compute the FFT
15 %
16 - Y=circshift(ffty,[0,N/2]); % Now shift the spectrum to put zero frequency
17 % at the middle of the array
18 %
19 %Compute the vector of frequencies
20 - df=Fs/N;
21 - Fvec=df*[-N/2:N/2-1];
22 %
```

# Fourier Series

- We can express any periodic function of time as sums of sine and cosine functions.
- Let's start with an even function that is periodic with a period  $T$



We could approximate this square wave with a cosine wave of the same period  $T$  and amplitude.

$$a_1 \cos(2\pi f_0 t)$$

Oscillation frequency

$$f_0 = \frac{1}{T}$$

Cycles per second (Hz)

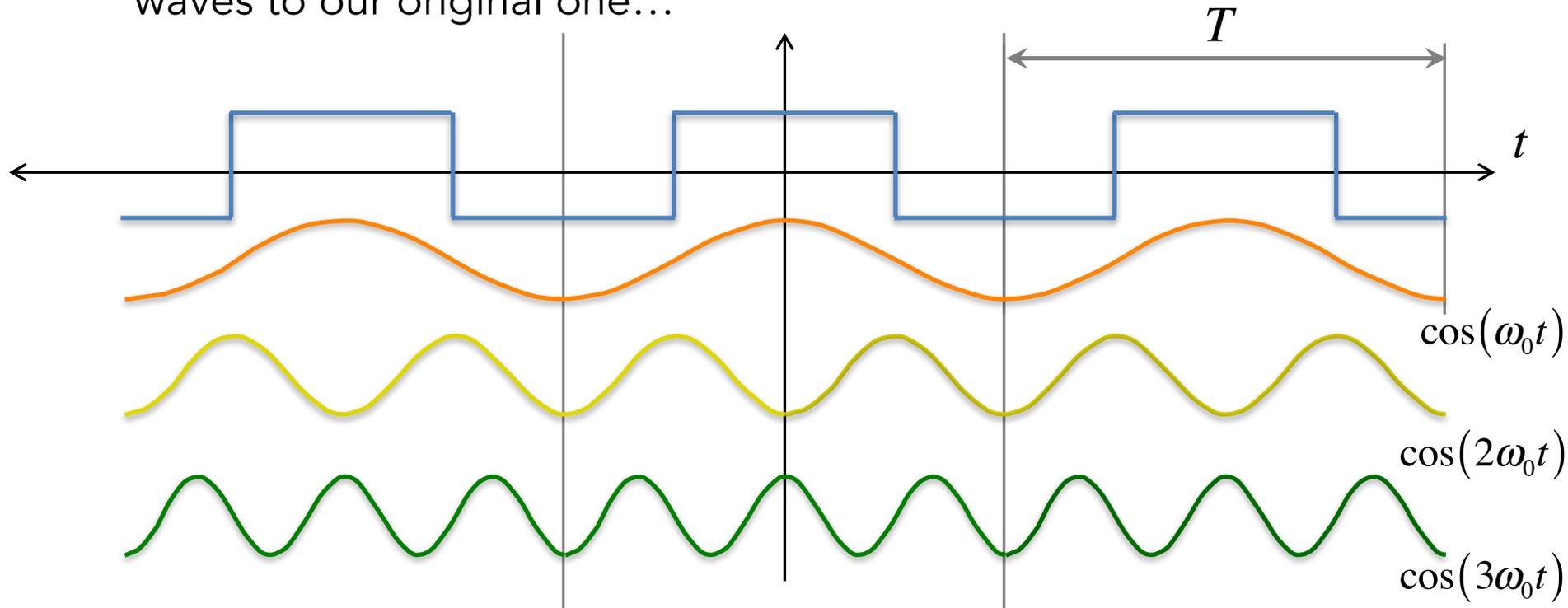
Angular frequency

$$\omega_0 = \frac{2\pi}{T}$$

Radians per second

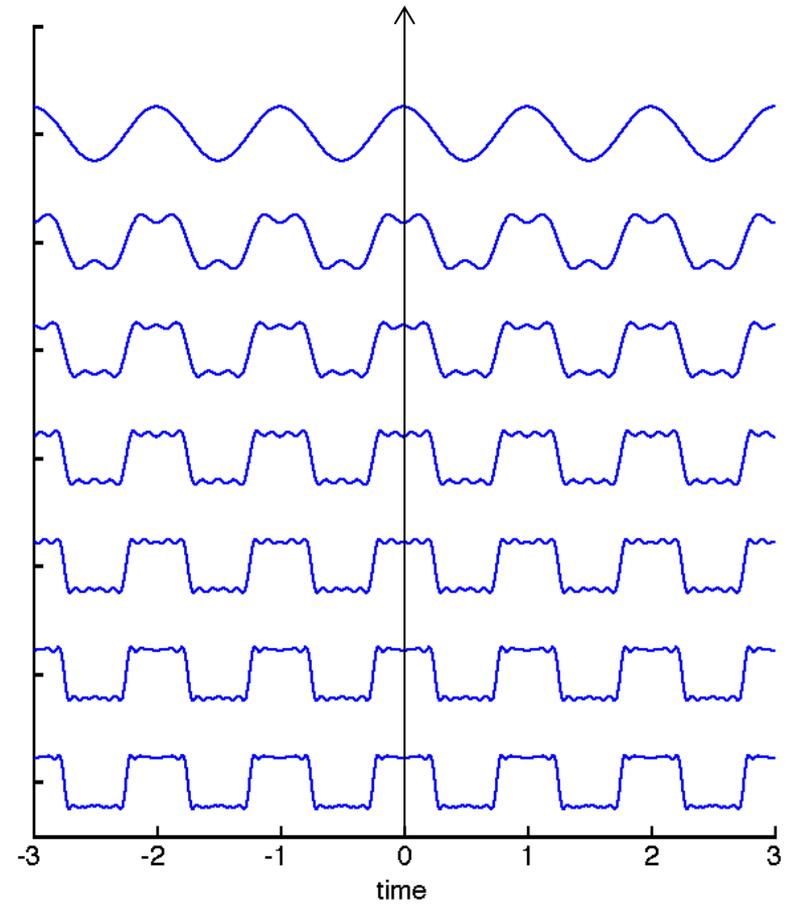
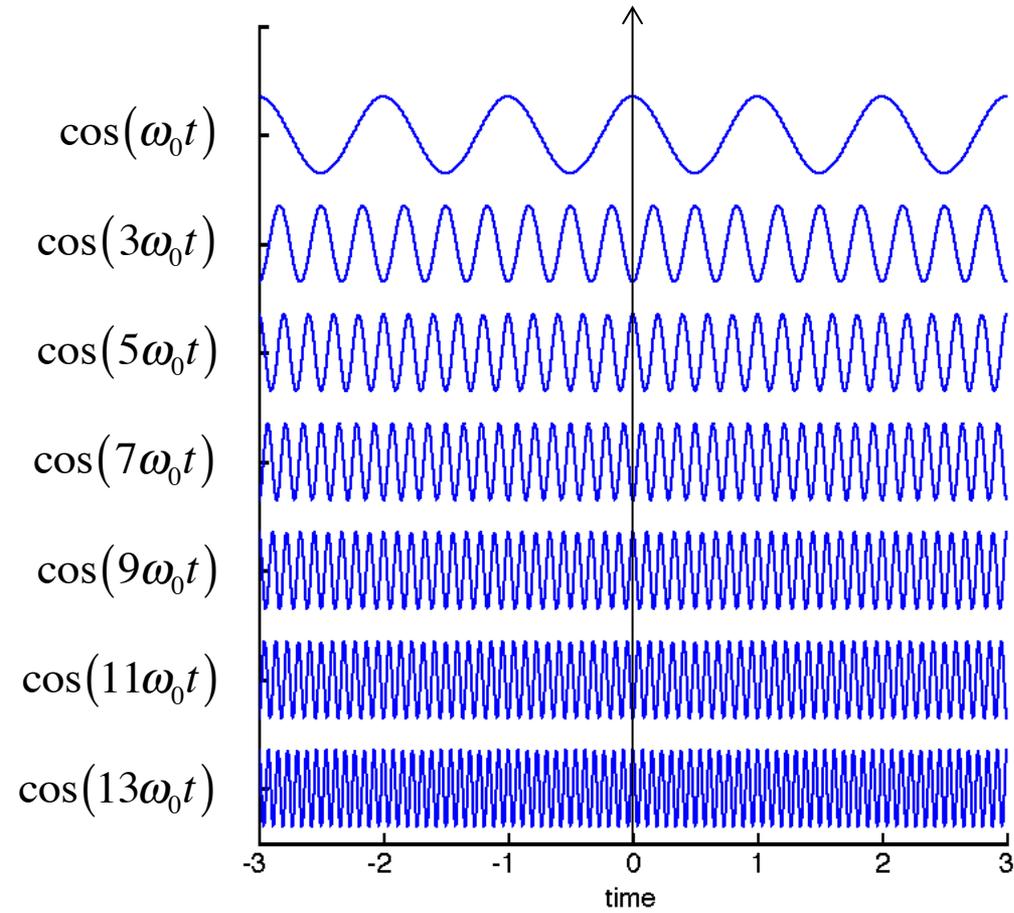
# Fourier Series

- But we can get a better approximation if we add some more cosine waves to our original one...



$$y(t) = a_1 \cos(\omega_0 t) + a_2 \cos(2\omega_0 t) + a_3 \cos(3\omega_0 t) + \dots$$

# Fourier Series

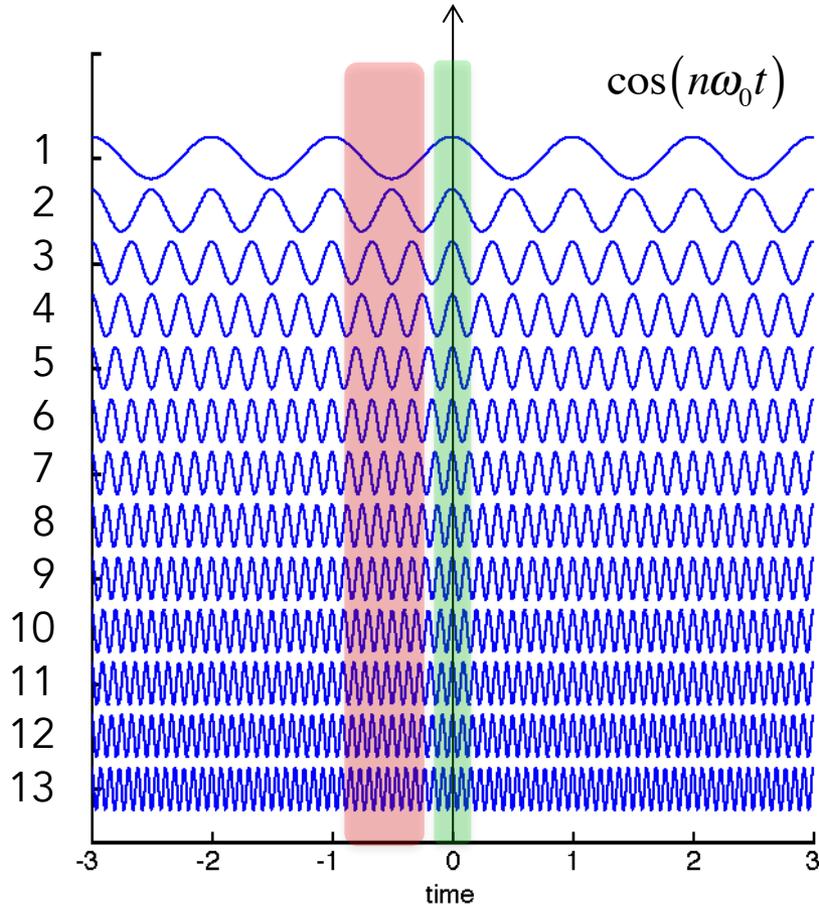


# Fourier Series

destructive  
interference

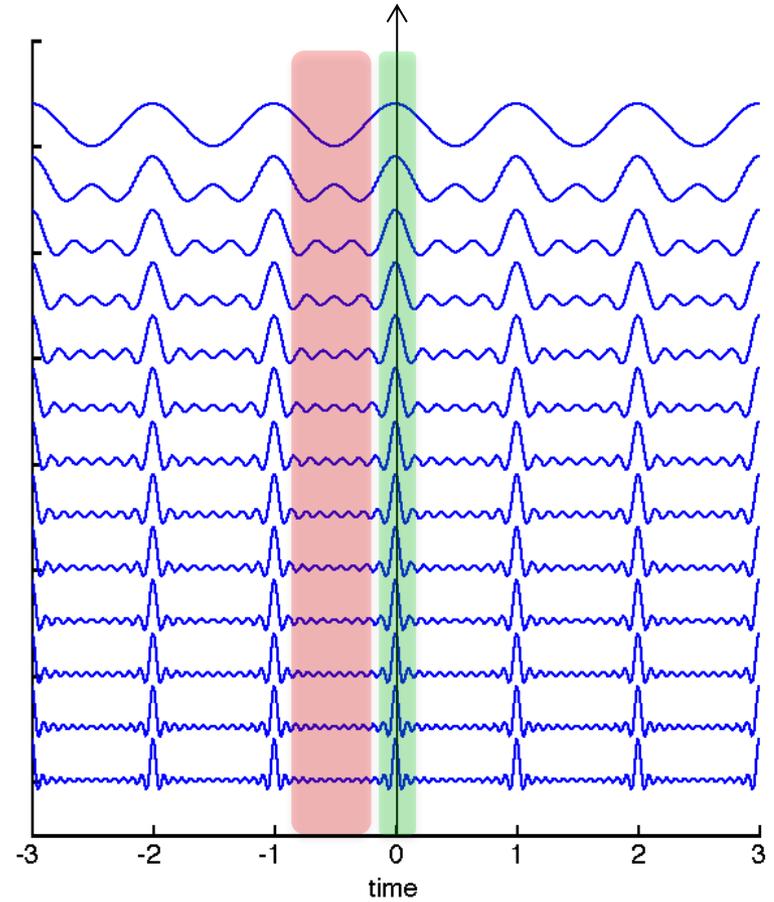
constructive  
interference

$n =$

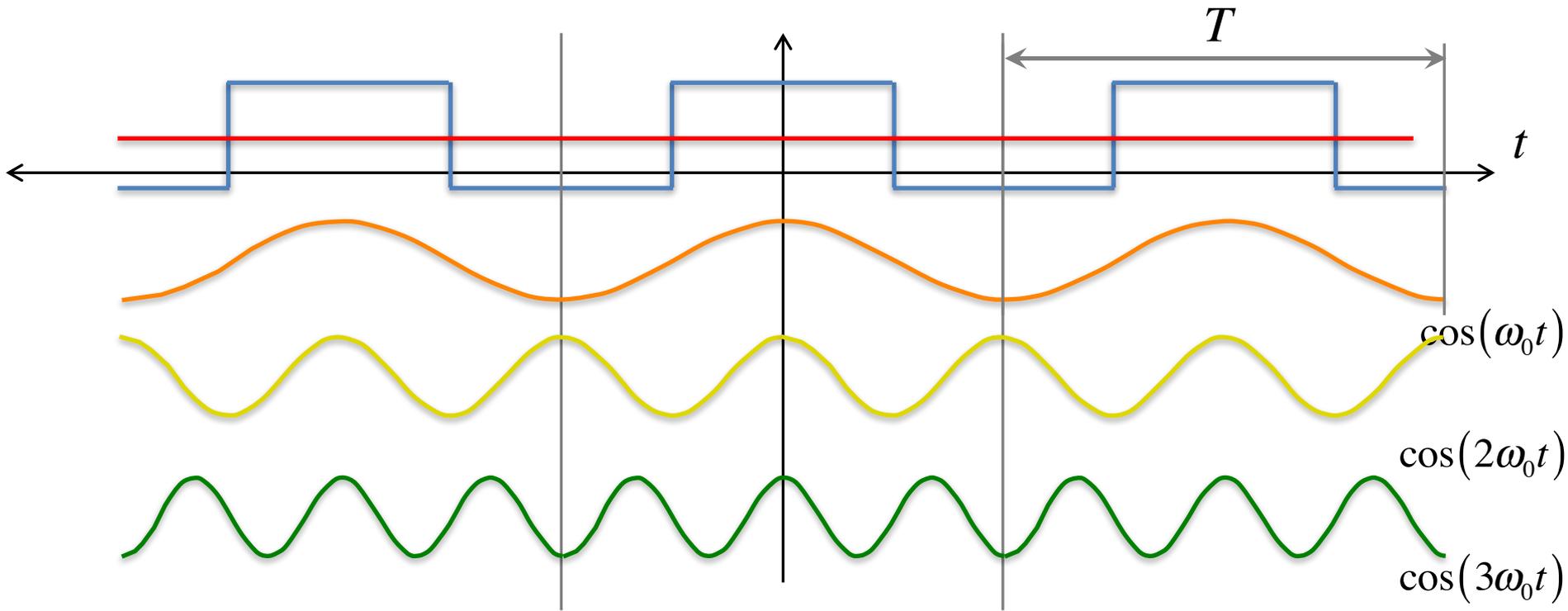


destructive  
interference

constructive  
interference



# Fourier Series



$$y(t) = \frac{a_0}{2} + a_1 \cos(\omega_0 t) + a_2 \cos(2\omega_0 t) + a_3 \cos(3\omega_0 t) + \dots$$

↑  
DC term

$$y_{\text{even}}(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} a_n \cos(n\omega_0 t)$$

# How do we find the coefficients?

- The  $a_0/2$  coefficient is just like the average of our function  $y(t)$ .

$$\frac{a_0}{2} = \frac{1}{T} \int_{-T/2}^{T/2} y(t) dt \quad a_0 = \frac{2}{T} \int_{-T/2}^{T/2} y(t) \cos(0\omega_0 t) dt$$

- The  $a_1$  coefficient is just the **overlap** of our function  $y(t)$  with  $\cos(\omega_0 t)$

$$a_1 = \frac{2}{T} \int_{-T/2}^{T/2} y(t) \cos(\omega_0 t) dt \quad \leftarrow \text{Correlation!}$$

- The  $a_2$  coefficient is just the overlap of our function  $y(t)$  with  $\cos(2\omega_0 t)$

$$a_2 = \frac{2}{T} \int_{-T/2}^{T/2} y(t) \cos(2\omega_0 t) dt$$

- The  $a_n$  coefficient is just the overlap of our function  $y(t)$  with  $\cos(n\omega_0 t)$

$$a_n = \frac{2}{T} \int_{-T/2}^{T/2} y(t) \cos(n\omega_0 t) dt$$

# How do we find the coefficients?

$$a_0 = \frac{2}{T} \int_{-T/2}^{T/2} y(t) dt$$

$$a_1 = \frac{2}{T} \int_{-T/2}^{T/2} y(t) \cos(\omega_0 t) dt$$

$$a_2 = \frac{2}{T} \int_{-T/2}^{T/2} y(t) \cos(2\omega_0 t) dt$$

Consider the following functions  $y(t)$ :

$$y(t) = 1$$

$$a_0 = 2$$

$$a_1 = 0$$

$$a_2 = 0$$

$$y(t) = \cos(\omega_0 t)$$

$$a_0 = 0$$

$$a_1 = 1$$

$$a_2 = 0$$

$$y(t) = \cos(2\omega_0 t)$$

$$a_0 = 0$$

$$a_1 = 0$$

$$a_2 = 1$$

$$\int_{-T/2}^{T/2} [\cos(\omega_0 t)]^2 dt = \frac{T}{2}$$

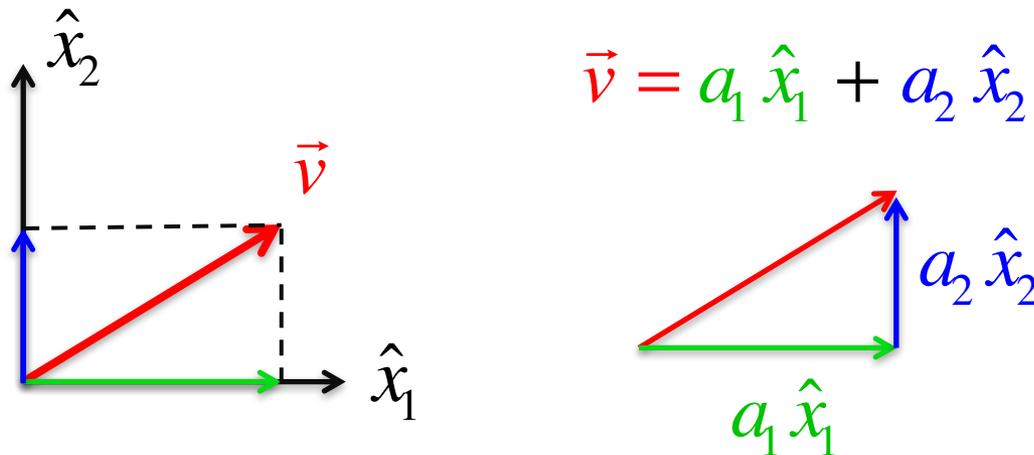
$$\int_{-T/2}^{T/2} \cos(\omega_0 t) \cos(2\omega_0 t) dt = 0$$

$$y(t) = \frac{a_0}{2} + a_1 \cos(\omega_0 t) + a_2 \cos(2\omega_0 t) + \dots$$

# Fourier Series

- If a function has maximal overlap with one of our cosine functions, then it has zero overlap with all the others!
- We say that our set of cosine functions form an orthogonal basis set...

$$u_n(t) = \cos(n\omega_0 t)$$



$$\vec{v} = a_1 \hat{x}_1 + a_2 \hat{x}_2$$

$$\hat{x}_1 = [0, 1]$$

$$\hat{x}_2 = [1, 0]$$

$$\vec{v} = [a_1, a_2]$$

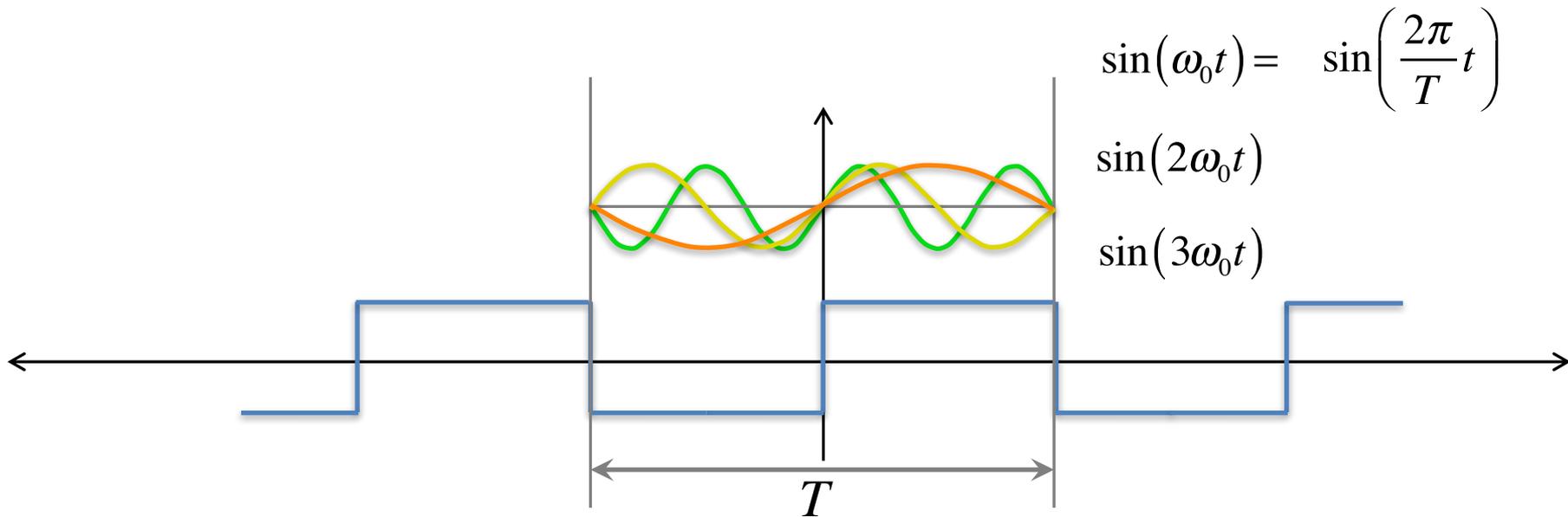
How do we find the coefficients  $a_1$  and  $a_2$ ?

$$a_1 = \vec{v} \cdot \hat{x}_1 = \sum_i v^i x_1^i$$

$$a_2 = \vec{v} \cdot \hat{x}_2 = \sum_i v^i x_2^i$$

# Fourier Series

- Now let's look at an odd (antisymmetric) function...



$$y_{odd}(t) = b_1 \sin(\omega_0 t) + b_2 \sin(2\omega_0 t) + b_3 \sin(3\omega_0 t) + \dots$$

$$y_{odd}(t) = \sum_{n=1}^{\infty} b_n \sin(n\omega_0 t)$$

Why is there no DC term here?

# Fourier Series

- For an arbitrary function, we can write it down as the sum of a symmetric and an antisymmetric part.

$$y(t) = \underbrace{\frac{a_0}{2} + \sum_{n=1}^{\infty} a_n \cos(n\omega_0 t)}_{\text{symmetric}} + \underbrace{\sum_{n=1}^{\infty} b_n \sin(n\omega_0 t)}_{\text{antisymmetric}}$$

# Learning Objectives for Lecture 11

- Fourier series for symmetric and asymmetric functions
- **Complex Fourier series**
- Fourier transform
- Discrete Fourier transform (Fast Fourier Transform - FFT)
- Power spectrum

# Complex Fourier Series

- We can express any periodic function of time as sums of complex exponentials.

Euler's formula

$$e^{i\omega t} = \cos \omega t + i \sin \omega t$$

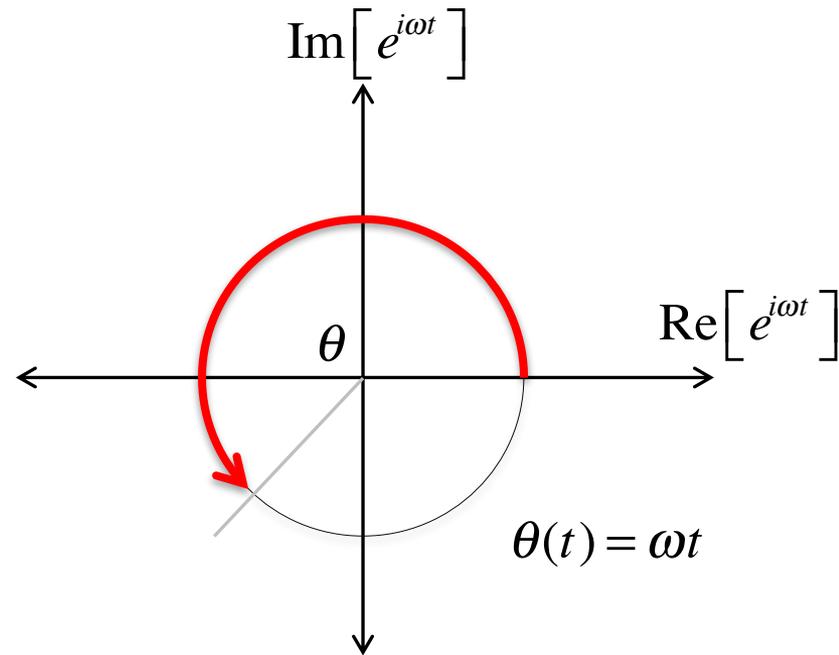
$$e^{-i\omega t} = \cos \omega t - i \sin \omega t$$

Rewrite as follows...

$$\cos \omega t = \frac{1}{2} (e^{i\omega t} + e^{-i\omega t})$$

$$\sin \omega t = \frac{1}{2i} (e^{i\omega t} - e^{-i\omega t}) = -\frac{i}{2} (e^{i\omega t} - e^{-i\omega t})$$

$$\frac{1}{i} = -i$$



# Fourier Series

$$y(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} a_n \cos(n\omega_0 t) + \sum_{n=1}^{\infty} b_n \sin(n\omega_0 t)$$

$$y(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} \frac{a_n}{2} (e^{in\omega_0 t} + e^{-in\omega_0 t}) + \sum_{n=1}^{\infty} \frac{-ib_n}{2} (e^{in\omega_0 t} - e^{-in\omega_0 t})$$

$$y(t) = A_0 + \sum_{n=1}^{\infty} A_n e^{in\omega_0 t} + \sum_{n=1}^{\infty} A_{-n} e^{-in\omega_0 t}$$

'DC' or  
'constant'  
term

positive  
frequencies

negative  
frequencies

$$A_0 = \frac{a_0}{2} \quad A_n = \frac{1}{2}(a_n - ib_n) \quad A_{-n} = \frac{1}{2}(a_n + ib_n) \quad A_n = (A_{-n})^*$$

complex conjugates

# Complex Fourier Series

$$y(t) = A_0 + \sum_{n=1}^{\infty} A_n e^{in\omega_0 t} + \sum_{n=1}^{\infty} A_{-n} e^{-in\omega_0 t}$$

- We can write this more compactly as follows:

$$= \sum_{n=0}^{\infty} A_n e^{in\omega_0 t} + \sum_{n=1}^{\infty} A_n e^{in\omega_0 t} + \sum_{n=-1}^{-\infty} A_n e^{in\omega_0 t}$$

For  $n = 0$ ,

$$e^{in\omega_0 t} = e^0 = 1$$

$$y(t) = \sum_{n=-\infty}^{\infty} A_n e^{in\omega_0 t}$$

# Complex Fourier Series

$$y(t) = \underbrace{\frac{a_0}{2} + \sum_{n=1}^{\infty} a_n \cos(n\omega_0 t)}_{\text{symmetric}} + \underbrace{\sum_{n=1}^{\infty} b_n \sin(n\omega_0 t)}_{\text{antisymmetric}}$$

- We can replace the sine and cosines of the fourier series with a single sum of complex exponentials

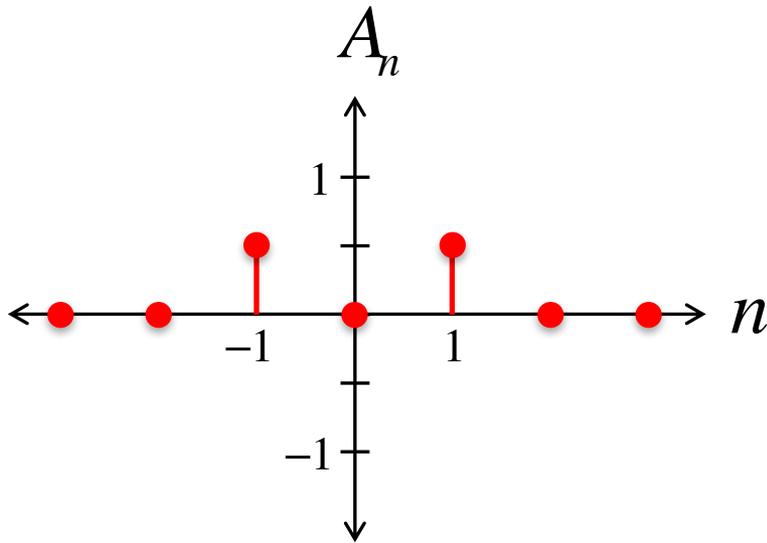
$$y(t) = \sum_{n=-\infty}^{\infty} A_n e^{in\omega_0 t}$$

# Complex Fourier Series

- Some examples...

$$A_{-1} = \frac{1}{2}, \quad A_0 = 0, \quad A_1 = \frac{1}{2}$$

$$y(t) = \sum_{n=-\infty}^{\infty} A_n e^{in\omega_0 t}$$



$$\cos \omega t = \frac{1}{2} (e^{i\omega t} + e^{-i\omega t})$$

$$y(t) = \frac{1}{2} e^{-i\omega_0 t} + \frac{1}{2} e^{i\omega_0 t} = \frac{1}{2} (\cos \omega_0 t - i \sin \omega_0 t) + \frac{1}{2} (\cos \omega_0 t + i \sin \omega_0 t)$$

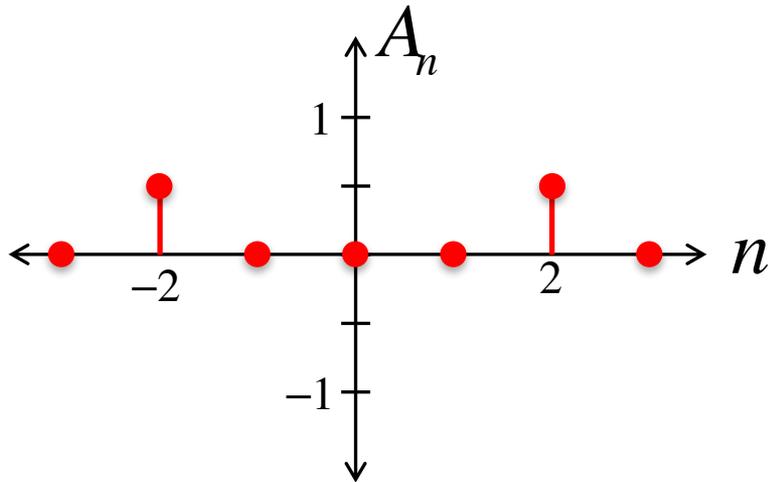
$$= \cos \omega_0 t$$

# Complex Fourier Series

- Some examples...

$$A_{-2} = \frac{1}{2}, \quad A_0 = 0, \quad A_2 = \frac{1}{2}$$

$$y(t) = \sum_{n=-\infty}^{\infty} A_n e^{in\omega_0 t}$$



$$y(t) = \frac{1}{2} e^{-i2\omega_0 t} + \frac{1}{2} e^{i2\omega_0 t}$$

$$= \frac{1}{2} (\cos 2\omega_0 t - i \sin 2\omega_0 t) + \frac{1}{2} (\cos 2\omega_0 t + i \sin 2\omega_0 t)$$

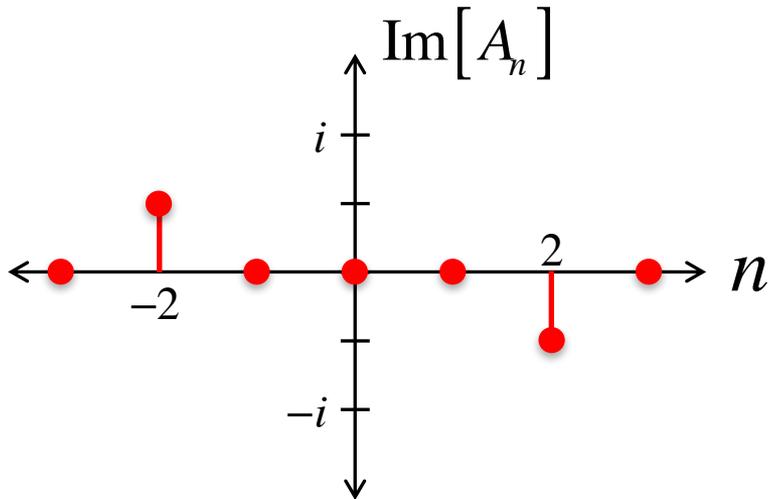
$$= \cos 2\omega_0 t$$

# Complex Fourier Series

- Some examples...

$$A_{-2} = \frac{i}{2}, \quad A_0 = 0, \quad A_2 = -\frac{i}{2}$$

$$y(t) = \sum_{n=-\infty}^{\infty} A_n e^{in\omega_0 t}$$



$$y(t) = \frac{i}{2} e^{-i2\omega_0 t} + \frac{-i}{2} e^{i2\omega_0 t}$$

$$= \frac{i}{2} (\cancel{\cos 2\omega_0 t} - i \sin 2\omega_0 t) + \frac{-i}{2} (\cancel{\cos 2\omega_0 t} + i \sin 2\omega_0 t) = \sin 2\omega_0 t$$

# Complex Fourier Series

- The set of functions  $e^{in\omega_0 t}$  form an orthogonal basis set over the interval  $\left[-\frac{T}{2}, \frac{T}{2}\right]$ .

- The  $A_0$  coefficient is just the average of our function  $y(t)$ .

$$A_0 = \frac{1}{T} \int_{-T/2}^{T/2} y(t) dt \qquad A_0 = \frac{1}{T} \int_{-T/2}^{T/2} y(t) e^{-0i\omega_0 t} dt$$

- The  $A_1$  coefficient is just the overlap of our function  $y(t)$  with  $e^{i\omega_0 t}$

$$A_1 = \frac{1}{T} \int_{-T/2}^{T/2} y(t) e^{-i\omega_0 t} dt$$

In general

$$A_m = \frac{1}{T} \int_{-T/2}^{T/2} y(t) e^{-im\omega_0 t} dt$$

$$y(t) = \sum_{n=-\infty}^{\infty} A_n e^{in\omega_0 t}$$

# Learning Objectives for Lecture 11

- Fourier series for symmetric and asymmetric functions
- Complex Fourier series
- **Fourier transform (I just want you to see this...)**
- Discrete Fourier transform (Fast Fourier Transform - FFT)
- Power spectrum

# Fourier Transform

(for non-periodic functions)

$$A_m = \frac{1}{T} \int_{-T/2}^{T/2} y(t) e^{-im\omega_0 t} dt$$

$$y(t) = \sum_{n=-\infty}^{\infty} A_n e^{in\omega_0 t}$$

- We are going to do this by letting the period go to infinity!

$$T \rightarrow \infty, \quad \omega_0 = \frac{2\pi}{T} \rightarrow 0, \quad m\omega_0 \rightarrow \omega, \quad A_m \rightarrow Y(\omega)$$

Fourier Transform

$$Y(\omega) = \int_{-\infty}^{\infty} y(t) e^{-i\omega t} dt$$

Inverse Fourier Transform

$$y(t) = \int_{-\infty}^{\infty} Y(\omega) e^{i\omega t} \frac{d\omega}{2\pi}$$

# Fourier transform

$$Y(\omega) = \int_{-\infty}^{\infty} y(t)e^{-i\omega t} dt$$

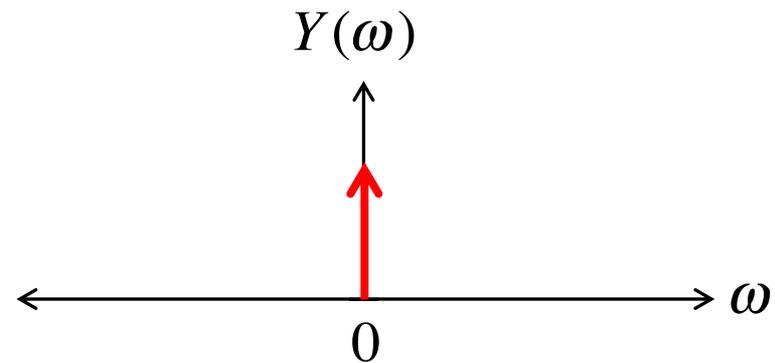
$$y(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} Y(\omega)e^{i\omega t} d\omega$$

- Some examples...

$$y(t) = 1$$

$$Y(\omega) = 2\pi\delta(\omega)$$

$$y(t) = \int_{-\infty}^{\infty} \delta(\omega)e^{j\omega t} d\omega = e^{i0t} = 1$$



# Fourier transform

$$Y(\omega) = \int_{-\infty}^{\infty} y(t) e^{-i\omega t} dt$$

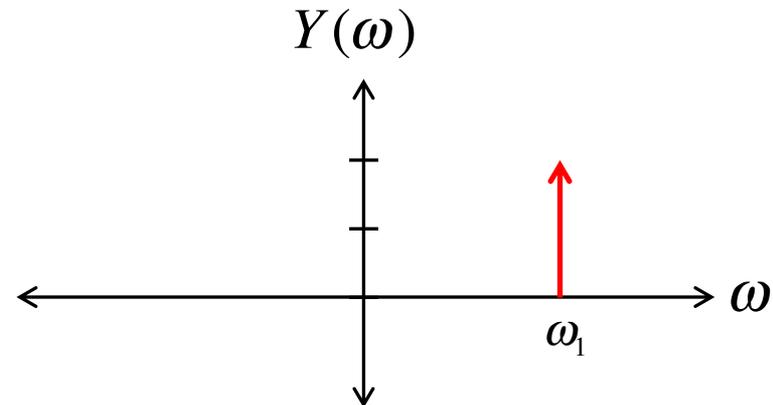
$$y(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} Y(\omega) e^{i\omega t} d\omega$$

- Some examples...

$$y(t) = e^{i\omega_1 t}$$

$$Y(\omega) = 2\pi \delta(\omega - \omega_1)$$

$$y(t) = \int_{-\infty}^{\infty} \delta(\omega - \omega_1) e^{i\omega t} d\omega = e^{i\omega_1 t}$$



# Fourier transform

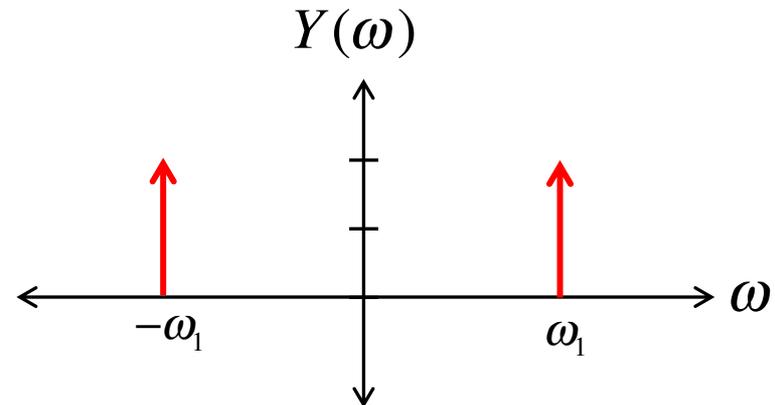
$$Y(\omega) = \int_{-\infty}^{\infty} y(t) e^{-i\omega t} dt$$

$$y(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} Y(\omega) e^{i\omega t} d\omega$$

- Some examples...

$$Y(\omega) = \pi [\delta(\omega + \omega_1) + \delta(\omega - \omega_1)]$$

$$y(t) = \frac{1}{2} e^{-i\omega_1 t} + \frac{1}{2} e^{i\omega_1 t} = \cos \omega_1 t$$



# Learning Objectives for Lecture 11

- Fourier series for symmetric and asymmetric functions
- Complex Fourier series
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- Discrete Fourier transform (Fast Fourier Transform - FFT)
- Power spectrum

# Discrete Fourier transform

- Computing the FT and IFT is, in principle really slow
- You have to compute an integral for every value of  $\omega$  you want in  $Y(\omega)$ .

$$Y(\omega) = \int_{-\infty}^{\infty} y(t) e^{-i\omega t} dt$$

- It turns out there is a \*super fast\* computer algorithm called the Fast Fourier Transform (FFT).

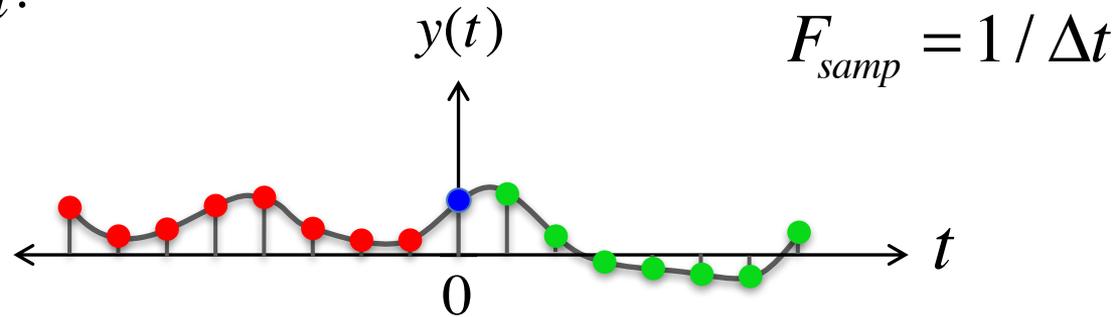
First, let's go back to oscillation frequency  $f$ , rather than angular frequency  $\omega$ :

$$f = \omega / 2\pi$$

$$Y(f) = \int_{-\infty}^{\infty} y(t) e^{-i2\pi f t} dt \qquad y(t) = \int_{-\infty}^{\infty} Y(f) e^{i2\pi f t} df$$

# Discrete Fourier Transform

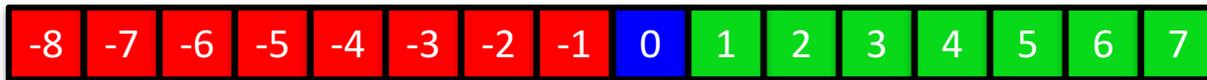
- Let's say we have a signal  $y(t)$  that is sampled at regular intervals  $\Delta t$ .



- Let's say we have  $N$  samples, and that  $N$  is an even number.

negative times

positive times



$$T_{\min} = -\frac{N}{2} \Delta t$$

$$T_{\max} = \left( \frac{N}{2} - 1 \right) \Delta t$$

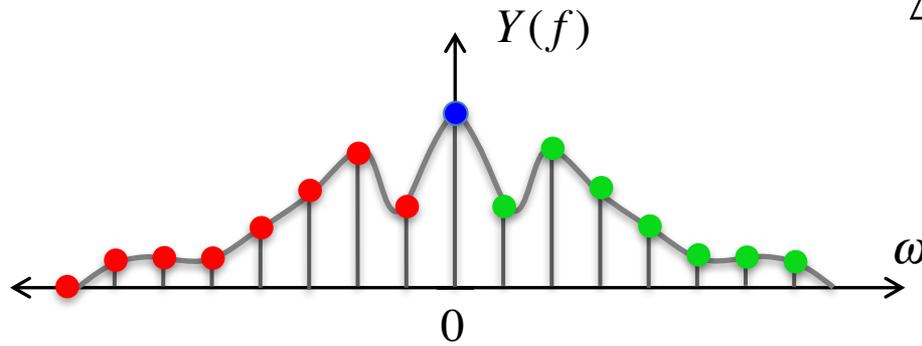
time=dt\*[-N/2:N/2-1]; % time array

# Discrete Fourier Transform

- The FFT algorithm returns a discrete Fourier transform that has  $N$  frequencies in frequency steps of  $\Delta f$

$$\Delta f = \frac{F_{s\text{amp}}}{N}$$

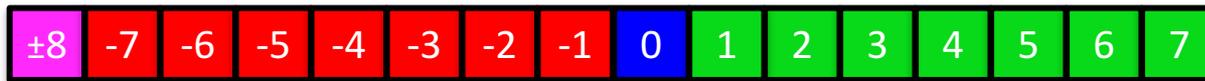
$$F_{\text{nyquist}} = \frac{F_{s\text{amp}}}{2}$$



constant term

negative frequencies

positive frequencies



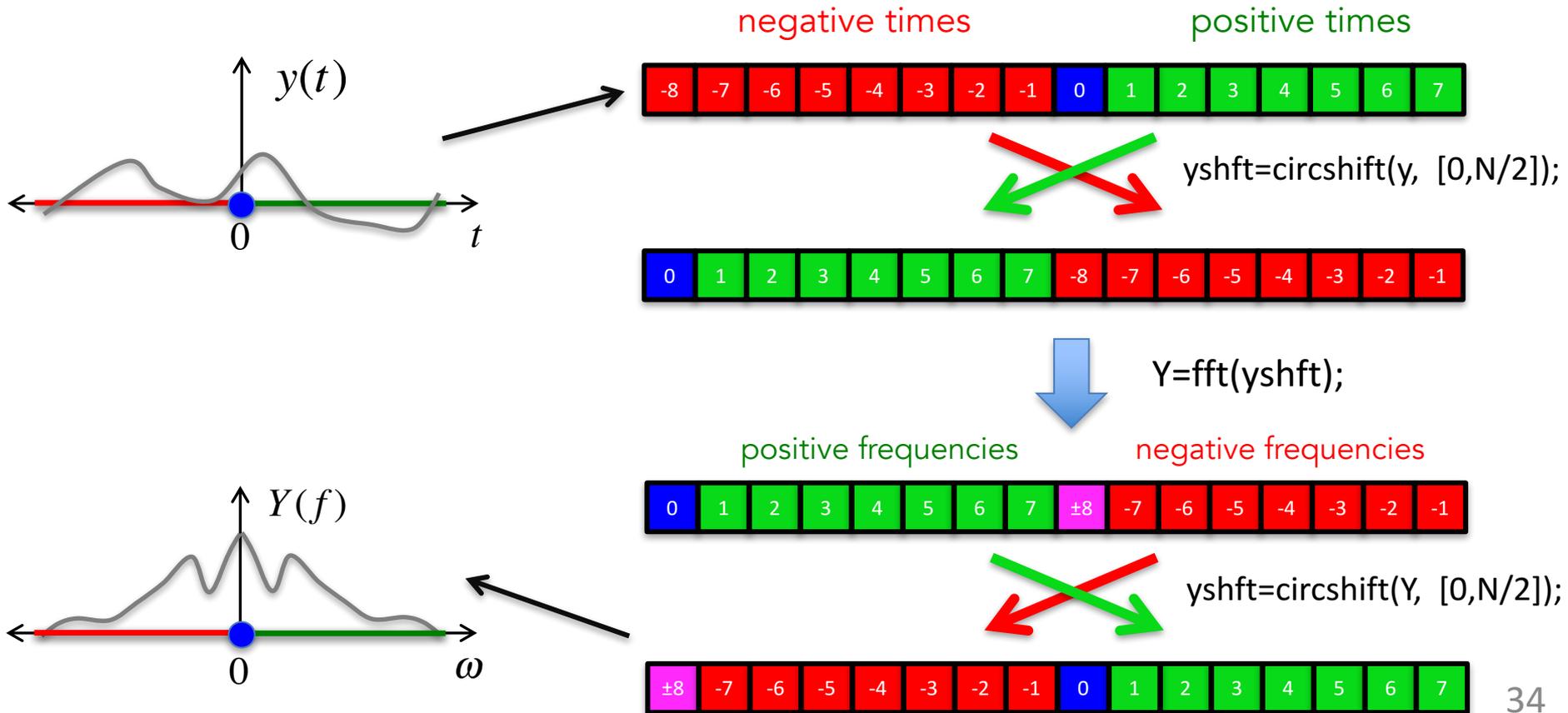
$$F_{\text{bottom}} = \pm \frac{N}{2} \Delta f$$

$$F_{\text{top}} = \left( \frac{N}{2} - 1 \right) \Delta f$$

freq=df\*[-N/2:N/2-1]; % frequency array

# Discrete Fourier transform

- One little trick... The FFT algorithm gets the time samples in a strange order, and returns the frequency samples in a strange order...



# Discrete Fourier transform

- Some code

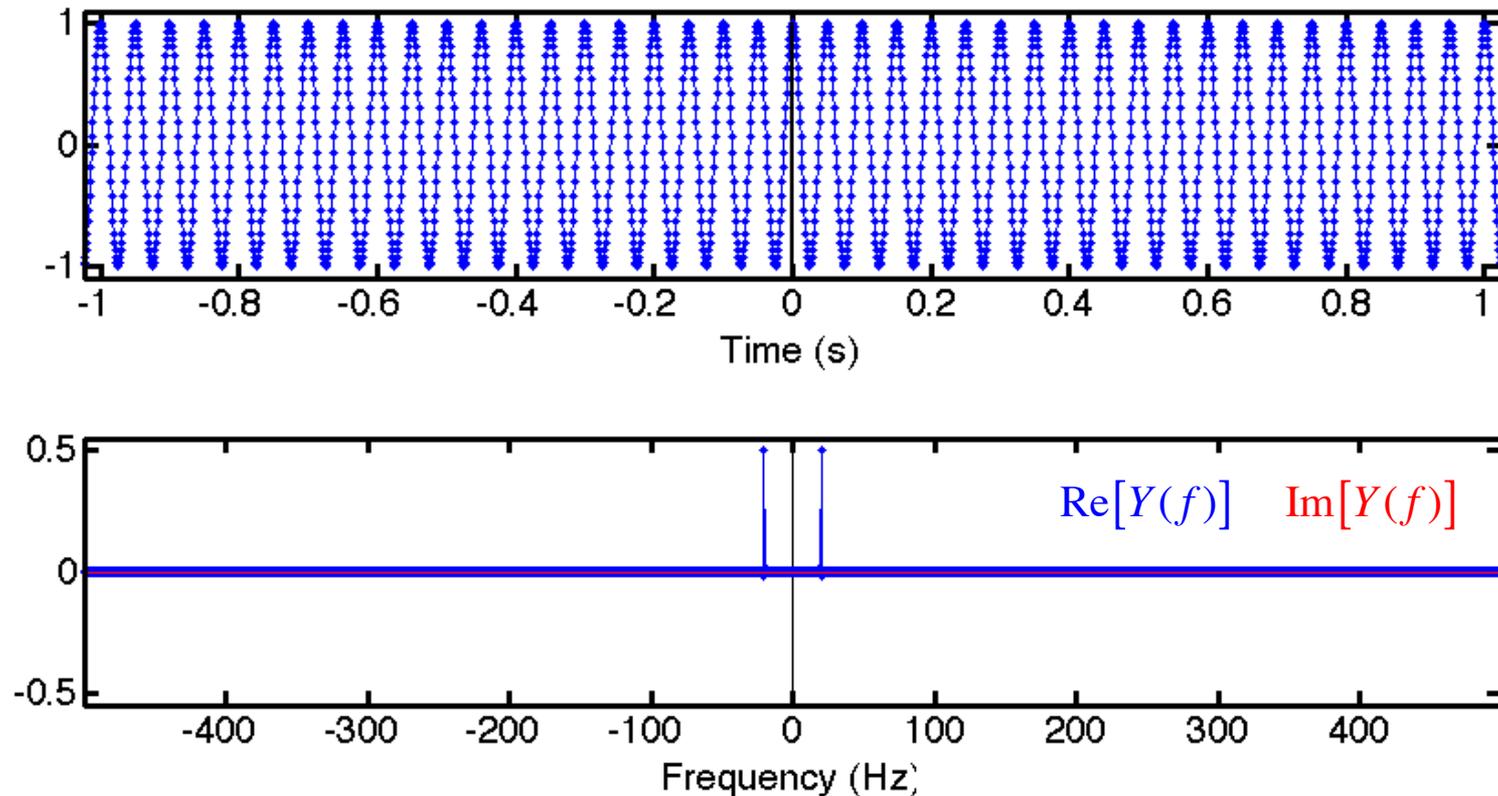
```
WSpec.m x recordaudio.m x continuous_cos.m x +
1 %
2 N=2048; % number of samples in time
3 %
4 dt=.001; % sampling interval
5 Fs=1./dt; % sampling frequency
6 time=dt*[-N/2:N/2-1]; % timebase
7 %
8 freq=20.; % frequency of sine wave in Hz
9 y=cos(2*pi*freq*time);
10 %
11 %
12 yshft=circshift(y,[0,N/2]); % First shift zero point from center to
13 % first point in the array
14 ffty=fft(yshft, N)/N; % Now compute the FFT
15
16 Y=circshift(ffty,[0,N/2]); % Now shift the spectrum to put zero frequency
17 % at the middle of the array
18 %
19 %Compute the vector of frequencies
20 df=Fs/N;
21 Fvec=df*[-N/2:N/2-1];
22 %
```

# Discrete Fourier transform

- Some examples – sine and cosine

$$y(t) = \cos(2\pi f_0 t) \quad f_0 = 20 \text{ Hz}$$

Continuous\_cos.m

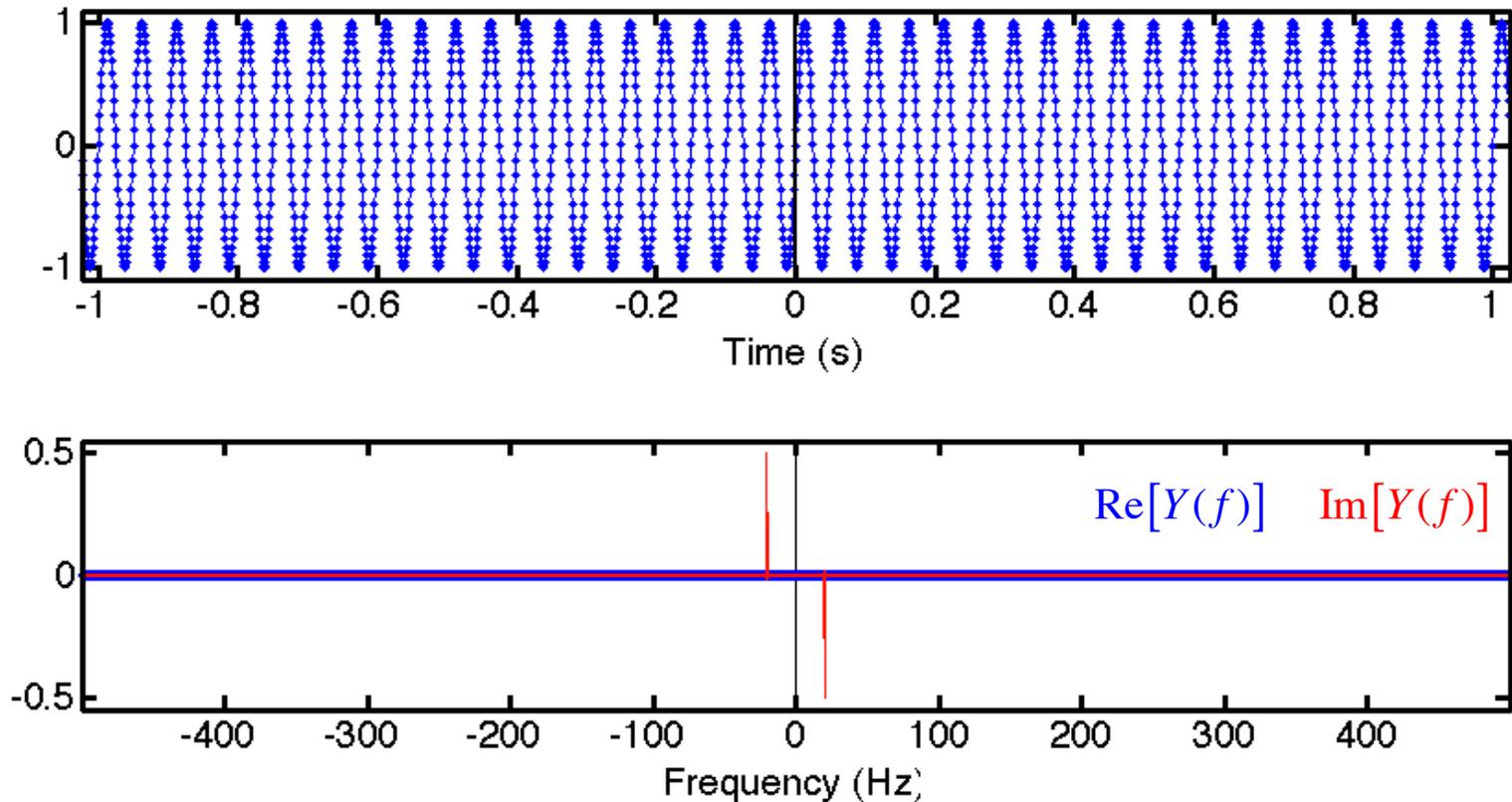


# Discrete Fourier transform

- Some examples – sine and cosine

$$y(t) = \sin(2\pi f_0 t) \quad f_0 = 20\text{Hz}$$

Continuous\_sin.m



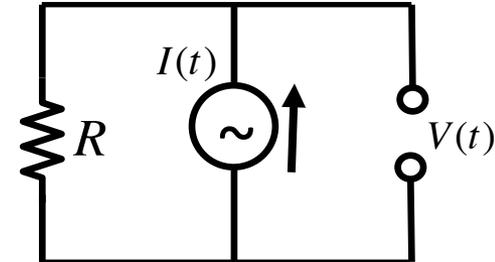
# Learning Objectives for Lecture 11

- Fourier series for symmetric and asymmetric functions
- Complex Fourier series
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- Discrete Fourier transform (Fast Fourier Transform - FFT)
- Power spectrum

# Introduce idea of ‘Power’

- The electrical power dissipated in a resistor is given by

$$P(t) = I(t)V(t) = \frac{1}{R}V^2(t)$$



- If the voltage is just a single sine wave at frequency  $\omega$ ...  $V(t) = \tilde{V}_\omega \cos(\omega t)$

$$V(t) = \tilde{V}_\omega \left[ \frac{1}{2} e^{-i\omega t} + \frac{1}{2} e^{i\omega t} \right]$$

Then the average power from one frequency component is just given by the square magnitude of the F.T. at that frequency...

$$P(\omega) = \frac{1}{R} |\tilde{V}_\omega|^2 \left( \left| \frac{1}{2} e^{-i\omega t} \right|^2 + \left| \frac{1}{2} e^{i\omega t} \right|^2 \right) = \frac{1}{R} |\tilde{V}_\omega|^2 \left( \left| \frac{1}{2} \right|^2 + \left| \frac{1}{2} \right|^2 \right) = \frac{1}{R} \frac{|\tilde{V}_\omega|^2}{2}$$

# Parseval's Theorem and Power

- The power in each frequency component independently contributes

$$E = \int_{-\infty}^{\infty} P(t) dt = \frac{1}{R} \int_{-\infty}^{\infty} [V(t)]^2 dt$$

Parseval's Theorem says that

$$\int_{-\infty}^{\infty} [V(t)]^2 dt = \int_{-\infty}^{\infty} |\tilde{V}(\omega)|^2 \frac{d\omega}{2\pi}$$

Power spectrum



Thus, each frequency component independently contributes to the power in the signal.

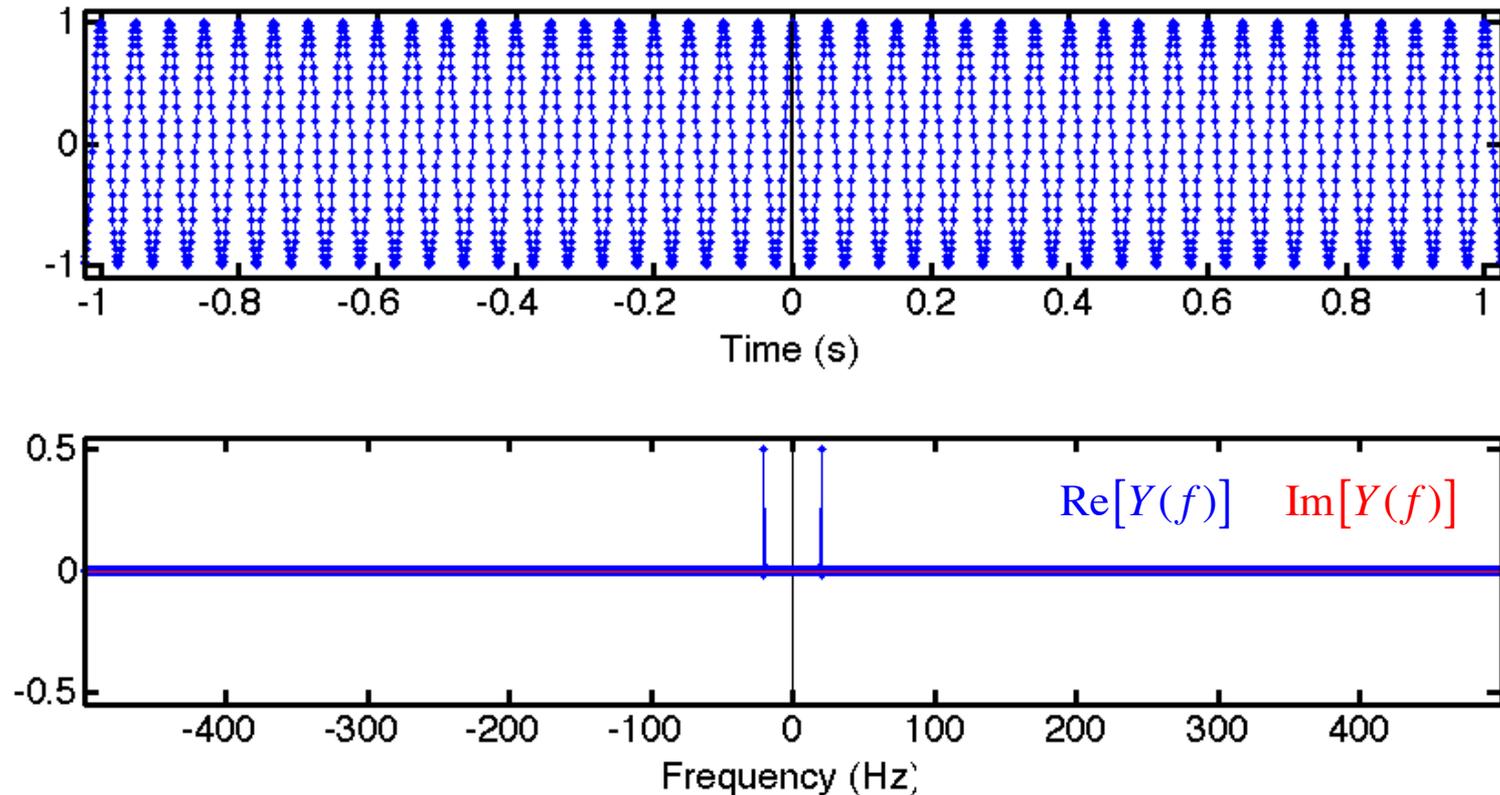
It also says that the total variance in the time domain signal is the same as the total variance in the frequency domain signal!

# Discrete Fourier transform

- Some examples – sine and cosine

$$y(t) = \cos(2\pi f_0 t) \quad f_0 = 20 \text{ Hz}$$

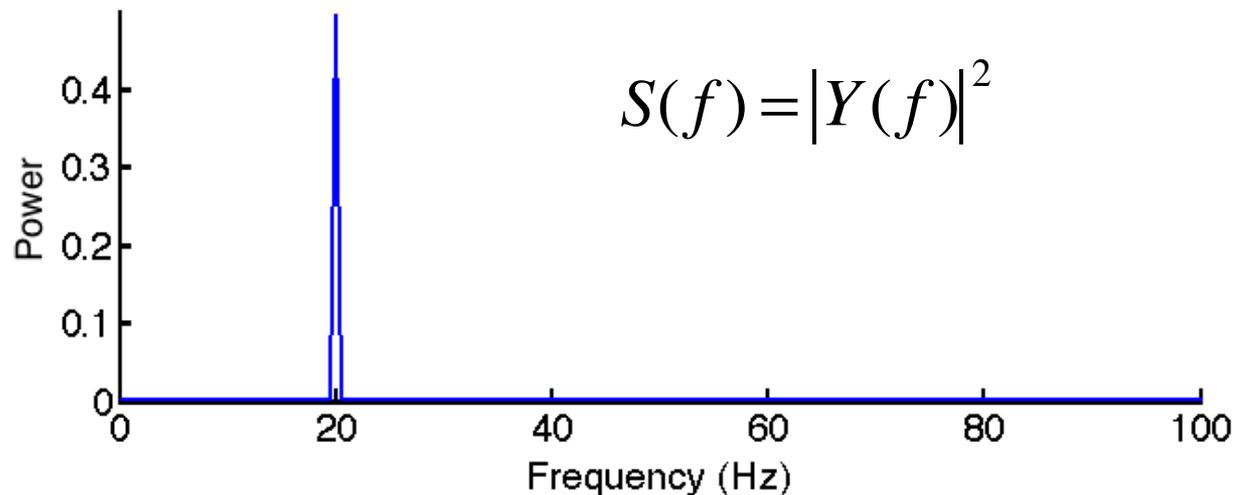
Continuous\_cos.m



# Discrete Fourier transform

- Power spectrum of sine and cosine

Continuous\_sin.m



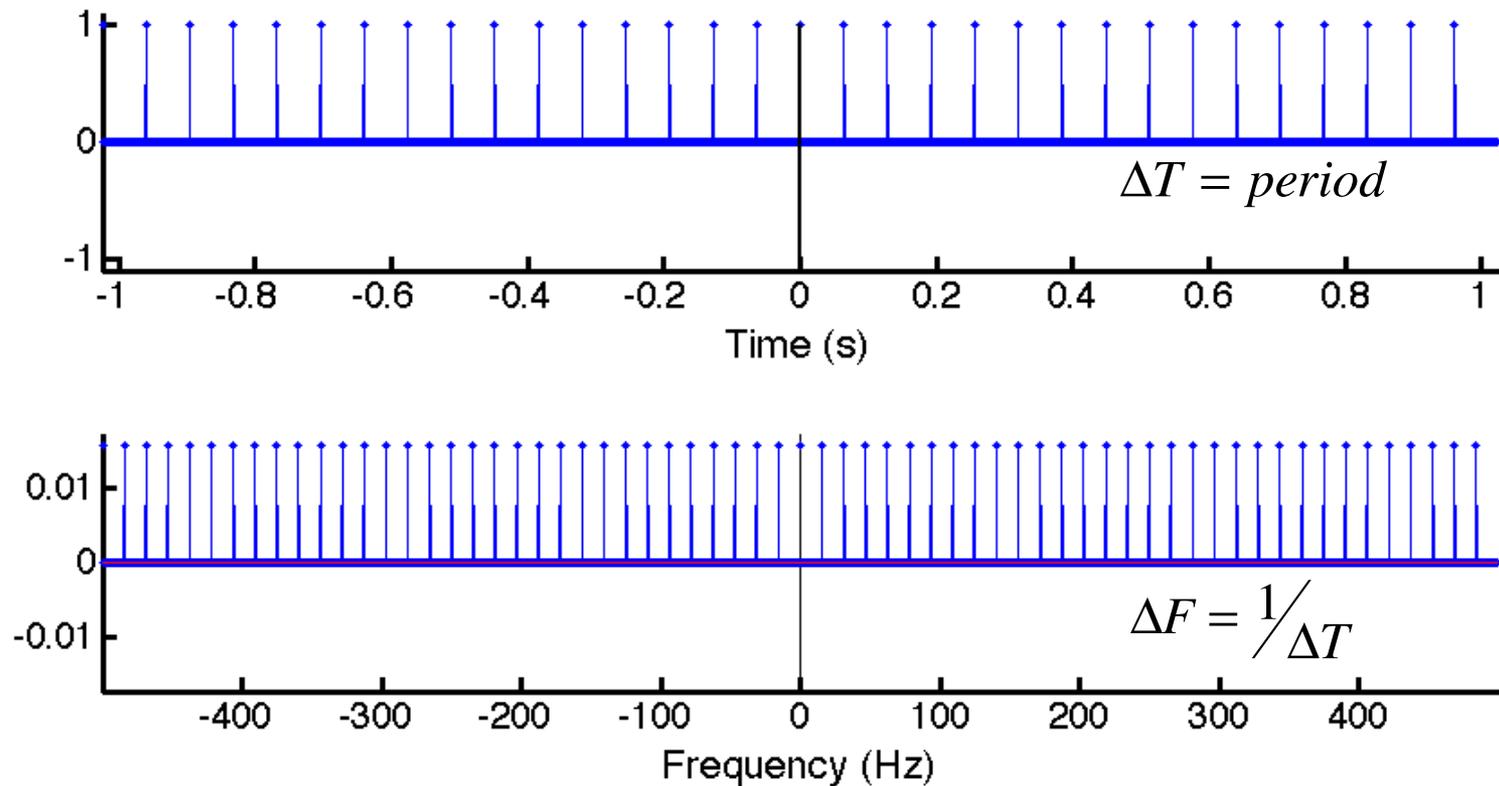
For real signals, the power spectrum is symmetric, so only need to plot for positive frequencies!

# Discrete Fourier transform

- Some examples – train of delta functions



deltafn\_train.m

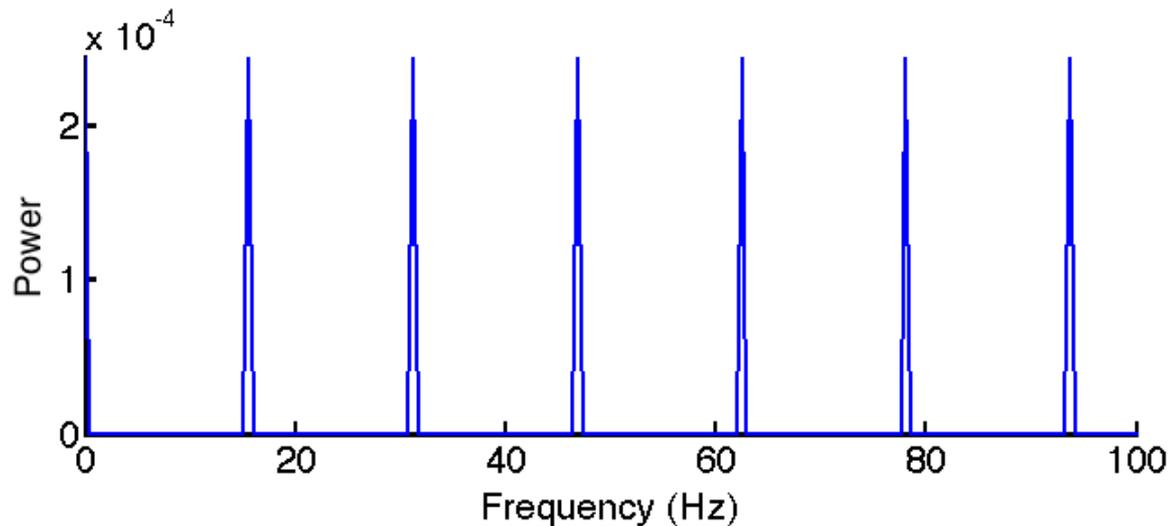


# Discrete Fourier transform

- Power spectrum– train of delta functions

$$S(f) = |Y(f)|^2$$

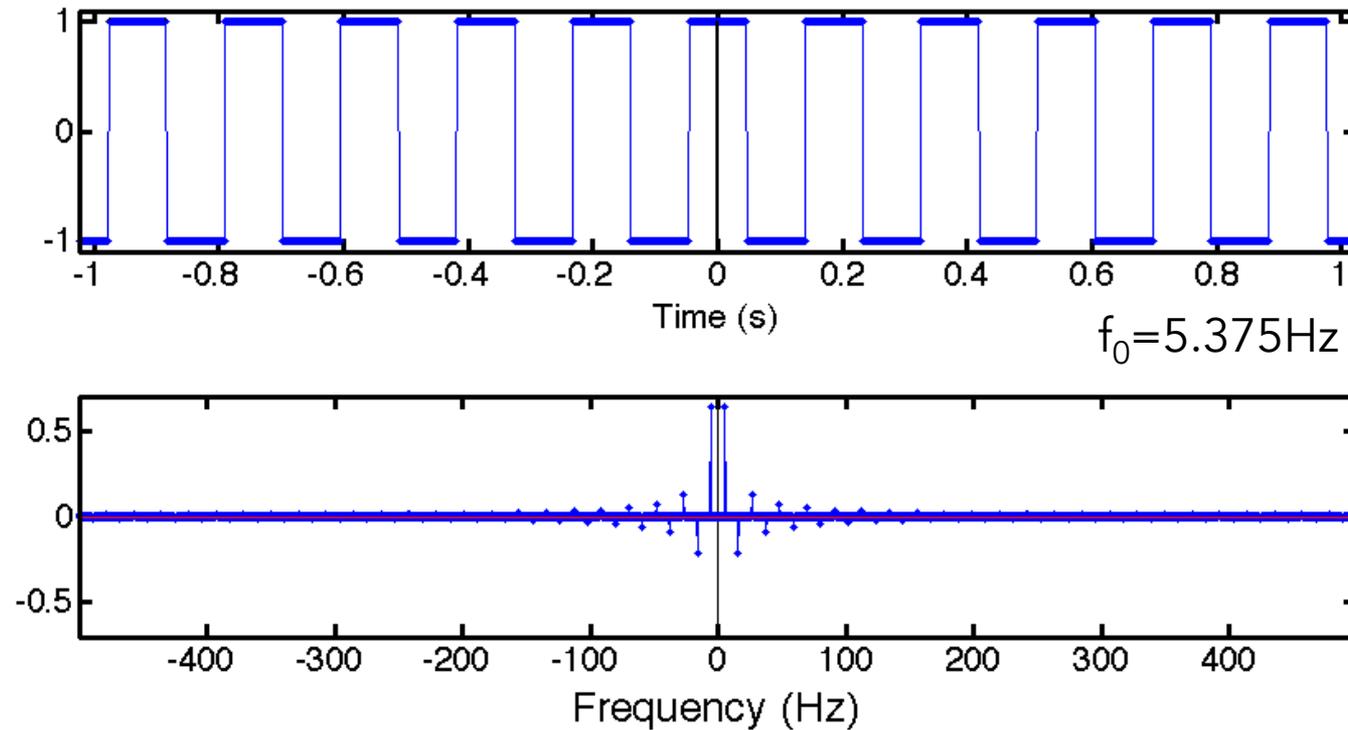
deltafn\_train.m



# Discrete Fourier transform

- Some examples – square waves

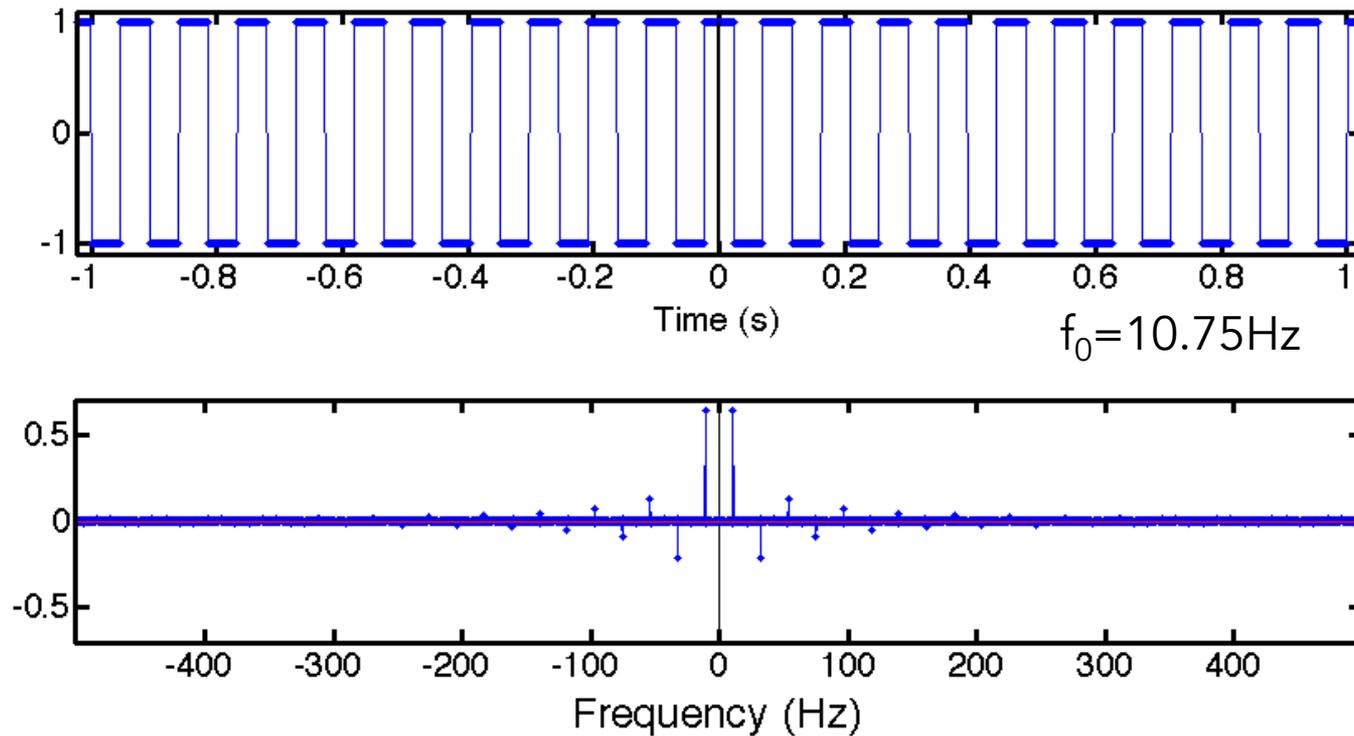
Continuous\_square.m



# Discrete Fourier transform

- Some examples – square waves

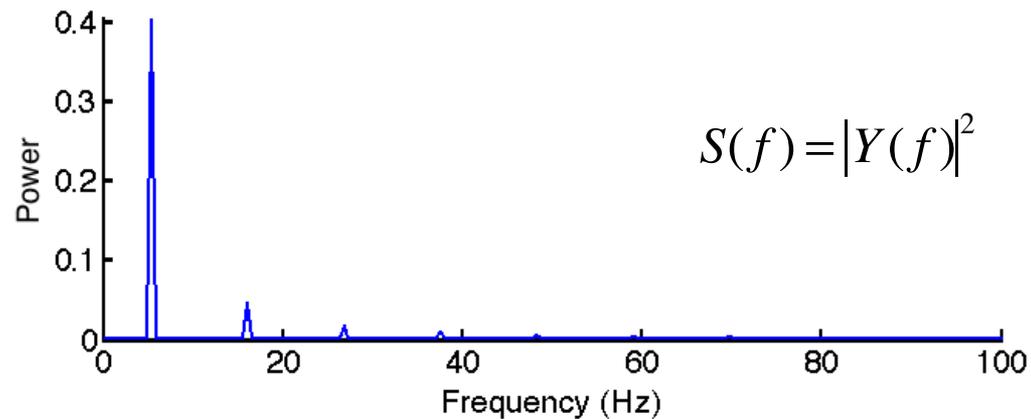
Continuous\_square.m



# Discrete Fourier transform

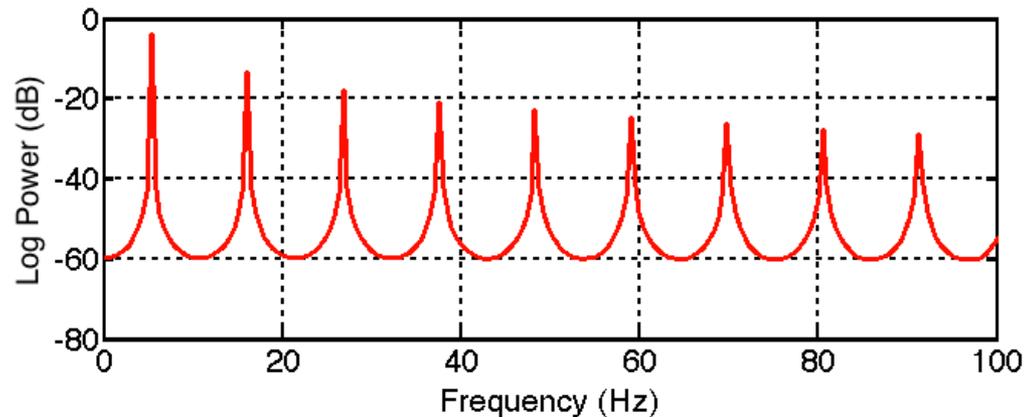
- Power spectrum– square wave

Continuous\_square.m



Spectrum plotted  
in units of  
decibels (dB)

$$10 \log_{10} S(f)$$



# Learning Objectives for Lecture 11

- Fourier series for symmetric and asymmetric functions
- Complex Fourier series
- Fourier transform
- Discrete Fourier transform (Fast Fourier Transform - FFT)
- Power spectrum

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9.40 Introduction to Neural Computation  
Spring 2018

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